

# INTERNATIONAL STANDARD

**IEC**  
**60489-2**

1991

AMENDMENT 1  
1999-07

---

---

## Amendment 1

### **Methods of measurement for radio equipment used in the mobile services –**

#### **Part 2: Transmitters employing A3E, F3E or G3E emissions**

## Amendement 1

### *Méthodes de mesure applicables au matériel de radiocommunication utilisé dans les services mobiles –*

#### *Partie 2: Émetteurs utilisant les émissions A3E, F3E ou G3E*

© IEC 1999 Copyright - all rights reserved

International Electrotechnical Commission 3, rue de Varembé Geneva, Switzerland  
Telefax: +41 22 919 0300 e-mail: [inmail@iec.ch](mailto:inmail@iec.ch) IEC web site <http://www.iec.ch>



Commission Electrotechnique Internationale  
International Electrotechnical Commission  
Международная Электротехническая Комиссия

PRICE CODE

**T**

*For price, see current catalogue*

## FOREWORD

This amendment has been prepared by IEC technical committee 102: Equipment used in radio communications for mobile services and for satellite communication systems.

The text of this amendment is based on the following documents:

FDIS	Report on voting
102/43/FDIS	102/52/RVD

Full information on the voting for the approval of this amendment can be found in the report on voting indicated in the above table.

A bilingual version of this amendment may be issued at a later date.

\_\_\_\_\_

Page 3

## CONTENTS

*Insert, after clause 12, the following new clause 13:*

13 Modulation sensitivity.....

*Renumber the existing clause 13 and amend its title as follows:*

14 Total distortion factor.....

*Renumber clause 14 as clause 15.*

*Insert the following new clause 16.*

16 Compressor characteristics .....

*Renumber the existing clauses 15, 16, 17 and 18, as 17, 18, 19 and 20, respectively.*

## APPENDICES

*Delete appendices B and D.*

*Rename appendix C as appendix B.*

Page 5

## PREFACE

*Add, in the list of quoted publications, after 60244-6 (1976) the following:*

60244-7 (1979): *Part 7: Cabinet radiation at frequencies above 1 GHz*

*Add, under 60489-1 (1983) the following:*

Amendment 1 (1996)

Amendment 2 <sup>1)</sup>

*With regard to amendment 2, add the following footnote:*

*Add, under 60489-3 (1988) the following:*

ITU-T Recommendation O.41 (1994): Psophometer for use on telephone-type circuits

Page 7

## SECTION 1: GENERAL, SUPPLEMENTARY DEFINITIONS AND CONDITIONS OF MEASUREMENT

### 5.4 Limitation of the measuring audio-frequency band

*Replace, on page 11, this subclause by the following:*

Because some properties, for example noise and audio-frequency harmonic distortion, depend upon the audio-frequency bandwidth of the test equipment, reproducible results can be obtained only when the band of audio-frequencies occupied by the demodulated signal is restricted to specified limits.

This restriction may be accomplished by means of a band-limiting filter preceding any audio-frequency measuring device and adapted to the type of signals to be transmitted. The filter may be incorporated within the measuring equipment. When measuring residual hum and noise, only the low-pass portion of the filter need be specified (see appendix A).

Page 13

## SECTION 2: METHODS OF MEASUREMENT

### 8.3.2 Method of measurement at the antenna terminals

*Add, on page 15, the following sentence to 8.3.2 a):*

If the band rejection filter exhibits unwanted attenuation (certain types of band rejection filters, such as those constructed with transmission lines or cavities, can have unwanted attenuation at odd multiples of the centre frequency of the rejection bandwidth), for the measurement of spurious components at and above the second harmonic of the carrier frequency, the band rejection filter shall be replaced by a high-pass filter whose cut-off frequency is approximately 1,5 times the carrier frequency.

---

<sup>1)</sup> To be published.

Page 17

*Replace, in figure 2, the text of legend 8 by the following:*

band-rejection filter or high-pass filter (if needed)

Page 19

### 8.4.3 Method of measurement

*Add, at the end of 8.4.3 e), the following note:*

NOTE – The permissible input level which does not degrade the receiver performance can be calculated from

$$P_{rt} = P_t - A \leq P_r + S - 10 \text{ dBm}$$

where

$P_{rt}$  is the permissible receiver input level, in decibels (dBm);

$P_t$  is the carrier power of transmitter under test, in decibels per metre (dBm);

$A$  is the attenuation recorded in step e), in decibels (dB);

$P_r$  is the reference sensitivity of the test receiver, in decibels per metre (dBm);

$S$  is the selectivity of the receiver, in decibels (dB).

Page 29

*Add, before 8.5.7 a), the following note 1:*

NOTE 1 – This method of measurement can be applied to the transmitter that has a modulating signal which is superimposed on the speech signal and cannot be turned off (e.g. continuous tone controlled squelch systems).

*Add, to 8.5.7 b), the following note 2:*

NOTE 2 – In the case of transmitters that have a modulating signal which is superimposed on the speech signal and cannot be turned off (e.g. continuous tone controlled squelch systems), operate the transmitter in modulated condition at the power level specified by the manufacturer (which may correspond to the carrier power measured as described in 8.2).

Page 33

## 9 Radiated radio-frequency power

*Replace the text of this clause by the following:*

These measurements are usually made only on transmitters having integral antennas.

### 9.1 General

The radiated radio-frequency power output of a transmitter may contain

- average radiated carrier power (see 9.2);
- maximum effective radiated carrier power (see 9.3);
- RFM (random field measurement) radiated carrier power (see 9.4);
- RFCD radiated carrier power (see 9.5);
- modulation components determining transmission quality (see clauses 12, 13, 14 and 16);

- other modulation components (e.g. components contributing to the adjacent channel power) (see 9.8);
- spurious narrow-bandwidth components (see 9.6);
- noise, both inside the band (see clause 18) and outside the band, where it is called spurious transmitter noise (see 9.7);
- inter-transmitter intermodulation products (see clause 11), and
- cabinet radiation (see 9.9).

The measured levels may be due to the radiation from the antenna, audio lines, control lines, power mains or from the cabinet. These measurements generally require the use of a test site or an RFCD. Guides for the use of such test sites and RFCDs are given in IEC 60489-1, annex A.

## 9.2 Average radiated carrier power for transmitters with integral antennas

### 9.2.1 Definition

Average of the radiated carrier powers in eight directions distributed at 45° angles in the horizontal plane.

NOTE – The average radiated carrier power can be measured in OATS (open area test sites) or in LRTS (low reflection test sites) or in AC (anechoic chambers). Measured values in OATS may differ from the values measured in the other two test sites because of the ground reflected wave effect.

### 9.2.2 Method of measurement

- a) Choose the test site and the measuring distance suitable for the frequency, the environmental conditions, the required measurement error and transmitter dimensions, from those described in annex A of IEC 60489-1.
- b) Place the equipment under test as illustrated in a subclause for the chosen test site in the above-mentioned annex.
- c) Measure the normal direction radiated power by the basic measuring procedure described in the subclause.

NOTE – The normal direction of the equipment under test is usually the operation side and is specified by the manufacturer. The normal direction may be the specific direction used for RFCD calibration measurements.

- d) Rotate the equipment under test 45° clockwise and measure this direction radiated power by the same procedure as in step c).
- e) Repeat step d) until values have been obtained for eight azimuth positions.

### 9.2.3 Presentation of results

Calculate the average of eight values measured in step d). Record measurement conditions of the test site.

## 9.3 Maximum effective radiated carrier power

### 9.3.1 Maximum effective radiated carrier power in the horizontal plane

#### 9.3.1.1 Definition

Maximum effective radiated carrier power in the horizontal plane.

NOTE – The maximum radiated carrier power can be measured in OATS, LRTS or AC. Measured values in OATS may differ from the values measured in the other two test sites because of the ground reflected wave effect.

### 9.3.1.2 Method of measurement for the maximum effective radiated carrier power in the horizontal plane

- a) Choose the test site and the measuring distance suitable for the frequency, the environmental conditions, the required measurement error and transmitter dimensions, from those described in annex A of IEC 60489-1.
- b) Place the transmitter under test as illustrated in a subclause for the chosen test site in the above-mentioned annex. Calibrate the test site.
- c) Measure the normal direction radiated power by the basic measuring procedure for radiation measurement described in the above-mentioned subclause.

NOTE – The normal direction of the equipment under test is usually the operation side and is specified by the manufacturer. The normal direction may be the specific direction used for RFCD calibration measurements.

- d) Rotate the transmitter under test in the horizontal plane and find the direction which has the maximum indication on the power measuring device. Note the maximum indication.

### 9.3.1.3 Presentation of results

Calculate the maximum effective radiated carrier power in the horizontal plane and record the direction in step d).

## 9.3.2 Maximum effective radiated carrier power on a spherical surface

### 9.3.2.1 Definition

Maximum value of effective radiated carrier power observed on a closed surface surrounding an antenna.

NOTE – The maximum radiated carrier power can be measured in LRTS or in AC.

### 9.3.2.2 Method of measurement for the maximum effective radiated carrier power on a spherical surface

Follow the procedure described in items a) to c) of 9.3.1.2.

- d) Rotate the transmitter under test in the horizontal plane and find the direction which has the maximum indication on the selective measuring device.
- e) Rotate the transmitter under test 90° in the vertical plane keeping the direction found in step d) to the measuring antenna.
- f) Orient the measuring antenna so that it is horizontally polarized. Rotate the transmitter under test in the horizontal plane and note the maximum indication. This will be the direction which has the maximum radiated carrier power on a spherical surface.
- g) If another direction offers more radiated carrier power, confirm it for that direction by repeating step d) to step f).

### 9.3.2.3 Presentation of results

Calculate the maximum effective radiated carrier power in step f) and record the direction in step d) as the horizontal plane direction and the direction in step f) as the vertical plane direction.

## 9.4 RFM (random field measurement) radiated carrier power

### 9.4.1 Definition

Median of radiated carrier power measurements taken along a random path in a random field.

NOTE – The RFM radiated carrier power can be measured in the RFM site.

#### 9.4.2 Method of measurement

- a) Confirm the construction parameter of the RFM site and position the transmitter under test as illustrated in annex A of IEC 60489-1.
- b) Measure the RFM radiated carrier power by the basic measuring procedure described in the above-mentioned annex.

#### 9.4.3 Presentation of results

Record the RFM radiated carrier power, the path length and the DDD value measured in step b).

NOTE – The DDD value expresses the measuring error or repeatability in the test site evaluation measurement described in the above-mentioned annex.

### 9.5 RFCD (radio-frequency coupling device) radiated carrier power

#### 9.5.1 Definition

Signal level at the output of an RFCD, in which the transmitter under test operates, calibrated by the specific direction radiated carrier power in a test site.

#### 9.5.2 Method of measurement

- a) Choose the RFCD for the frequency and size of the transmitter under test from those described in annex A of IEC 60489-1.
- b) Calibrate or confirm the calibration value for the output signal level of the RFCD. The calibration should be performed with the calibration method for the chosen RFCD described in the subclause of the above-mentioned annex.
- c) Place the transmitter under test in the specified position and in the specified direction and activate the transmitter.
- d) Measure the output level of the RFCD.

#### 9.5.3 Presentation of results

Calculate the RFCD radiated carrier power from the relation of the values in step b) and step d). Record the position, the direction of the transmitter under test in the RFCD, and the calibration value.

### 9.6 Radiated spurious narrow-bandwidth components

The methods of measurement for the radiated spurious narrow-bandwidth components are principally the same as those of the radiated carrier power described in 9.2 to 9.5. The only difference for radiated spurious measurement is the presence of a very high-level carrier. Therefore, the measurement conditions are the same as the combination of the above radiated carrier power measurement and the measurement of the terminal spurious narrow-bandwidth radio-frequency components in 8.3.

The different kinds of radiated spurious narrow-bandwidth components are as follows:

- average radiated spurious narrow-bandwidth components;
- maximum radiated spurious narrow-bandwidth components;
- RFM radiated spurious narrow-bandwidth components;
- RFCD radiated spurious narrow-bandwidth components.

### 9.7 Radiated spurious transmitter noise

#### 9.7.1 Definition

Continuous spectrum of noise components radiated by transmitters with integral antennas.

NOTE – For more general information, see 8.4.1.

### 9.7.2 Method of measurement

This method uses the RFCD or near-field coupling of the transmitter and the test antenna in the Faraday cage (if necessary) and the output of the RFCD, or the test antenna is measured with the same method as for one of the terminal radio-frequency powers, for example 8.4. The method uses the ratio measurement, and the coupling of the RFCD or the coupling of the transmitter and the test antenna is calculated from the radiated carrier power measured as described in 9.2.

## 9.8 Radiated adjacent channel power

### 9.8.1 Definition

That part of the total power, under defined conditions of modulation, which falls within a specified bandwidth centred on the centre frequency of either of the adjacent channels radiated from transmitters with integral antennas.

### 9.8.2 Method of measurement

This method uses the RFCD or near-field coupling of the transmitter and the test antenna in the Faraday cage (if necessary), and the output of the RFCD or the test antenna is measured with the same method for one of the terminal radio-frequency powers, for example 8.5. The method uses the ratio measurement, and the coupling of the RFCD or the coupling of the transmitter and the test antenna is calculated from the radiated carrier power measured as described in 9.2 to 9.4.

## 9.9 Cabinet radiation

### 9.9.1 Definition

The cabinet radiation of transmitters with integral antennas is contained in the radiated spurious narrow-bandwidth components measured as described in 9.6.

### 9.9.2 Method of measurement

Methods of measurement for the cabinet radiation of transmitters equipped with suitable antenna terminals are given in IEC 60244-6 and IEC 60244-7.

Page 57

## 11 Inter-transmitter intermodulation

*Replace the text of this clause by the following:*

### 11.1 Definition

Process by which intermodulation products are generated in the output circuits of a transmitter due to the presence of an unwanted signal from another transmitter.

#### 11.1.1 Inter-transmitter intermodulation (ITIM)

For the purpose of this standard, inter-transmitter intermodulation (ITIM) is expressed in terms of the ratio, in decibels, of the transmitter carrier power to the power of the third-order intermodulation product caused by the presence of an interfering signal which is 30 dB below the level of the transmitter output and which is incident upon the output of the disturbed transmitter.



### 11.1.2 Transmitter intermodulation conversion loss (TIMCL)

For the purpose of this standard, transmitter intermodulation conversion loss (TIMCL) is expressed in terms of the ratio, in decibels, of the power of the interfering signal (which is 30 dB below the transmitter carrier power) which is incident upon the output of the disturbed transmitter to the power of the third-order intermodulation product.

NOTE – Inter-transmitter intermodulation (ITIM) and transmitter intermodulation conversion loss (TIMCL) are related arithmetically by the expression

$$\text{ITIM} = \text{TIMCL} + 30 \text{ dB}$$

## 11.2 Method of measurement (antenna terminals)

### 11.2.1 Method of measurement using a line stretcher and a circulator

a) Connect the equipment as shown in figure 8a.

NOTE 1 – The impedance of all components in the measuring arrangement (with the possible exception of the output impedance of the transmitter under test) should be the same as the characteristic impedance of the measuring arrangement transmission line. The directional couplers should be used to measure the incident and the reflected powers, and to make certain that there is not a gross mismatch between the test load and the measuring arrangement.

- b) Operate the transmitter without modulation at rated power output and adjust the selective voltmeter (3) for maximum indication at the operating frequency of the transmitter. Record the carrier level with the switch (4) in position A.
- c) Adjust the frequency of the radio-frequency test signal source (9) to a frequency 100 kHz above the operating frequency of the transmitter.
- d) Change switch (4) to position B. Adjust the selective voltmeter for maximum reading at the frequency of the radio-frequency test signal source (9) and adjust the test signal level to produce a value 30 dB below the level recorded in step b).
- e) With switch (4) in position A and the transmitter operating at rated power, adjust the selective voltmeter for maximum reading at the frequency of the third-order intermodulation product which is 100 kHz below the transmitter operating frequency.
- f) Adjust the line stretcher (5) until the maximum level of the intermodulation product is obtained.
- g) Repeat step d) and record the level of the signal from the radio-frequency test signal source.
- h) Repeat step e) and record the level of the intermodulation product.
- i) Change switch (4) to position B and note the reading of the selective voltmeter. A level less than 10 dB lower than that measured in step h) indicates a source of intermodulation other than the transmitter under test, for example the radio-frequency test signal source (9) or the circulator (7). This shall be eliminated and step h) shall then be repeated.
- j) The ratio, in decibels, of the level of the transmitter signal measured in step b) to the level of the intermodulation product measured in step h) is the inter-transmitter intermodulation (ITIM).
- k) The ratio, in decibels, of the level of the unwanted signal level measured in step g) to the level of the intermodulation product measured in step h) is the transmitter intermodulation conversion loss (TIMCL).

NOTE 2 – The method of measurement can be used for other frequency spacings and levels of the radio-frequency test signal source (9).

NOTE 3 – The selective voltmeter (3) should have a selectivity sufficiently high so as not to be influenced by the carrier level when measuring the intermodulation product.

### 11.2.2 Method of measurement using a resistive or a capacitive coupler

- a) Connect the equipment as shown in figure 8b.

NOTE 1 – In order to reduce mismatch errors, it is important that the 10 dB power attenuator (2) is coupled to the transmitter with the shortest possible connection.

- b) Operate the transmitter without modulation at rated power output and adjust the spectrum analyser (8) to give a maximum indication with a frequency span of 500 kHz.  
c) Adjust the frequency of the unmodulated radio-frequency test signal source (6) to a frequency 100 kHz above the operating frequency of the transmitter.  
d) Adjust the level of the value monitored on the spectrum analyser to 10 dB below the level of the transmitter (as displayed on the spectrum analyser).

NOTE 2 – This procedure will ensure that the ratio of the levels at the transmitter terminals is 30 dB.

- e) Measure the intermodulation component by direct observation on the spectrum analyser (8) and record the ratio, in decibels, of the level of the largest intermodulation component to the level of the interfering signal.  
f) Calculate the inter-transmitter intermodulation (ITIM) using the following relationship:

$$\text{ITIM} = X - 2 A_2 \text{ dB}$$

where

$X$  is the ratio recorded in step e);

$A_2$  is the attenuation of the power attenuator (2) (nominally 10 dB).

- g) Calculate the transmitter intermodulation conversion loss (TIMCL) using the following relationship:

$$\text{TIMCL} = X - 2 A_2 - 30 \text{ dB}$$

where

$X$  is the ratio recorded in step e);

$A_2$  is the attenuation of the power attenuator (2) (nominally 10 dB).

NOTE 3 – This method of measurement may be used for other frequency spacings and levels of radio-frequency test signal source (9).

NOTE 4 – The spectrum analyser should have a selectivity sufficiently high so as not to be influenced by the transmitter carrier level when measuring the intermodulation product.

### 11.3 Method of measurement (integral antenna)

#### 11.3.1 Method of measurement using a circulator and a line stretcher

- a) Connect the equipment as shown in figure 8c.

NOTE 1 – The impedance of all components in the measuring arrangement (with the possible exception of the output impedance of the transmitter) should be the same as the characteristic impedance of the measuring arrangement transmission line. The directional couplers should be used to measure the incident and reflected powers, and to make certain that there is not a gross mismatch between the test load and the measuring arrangement. The value of the maximum effective radiated power,  $P_{\text{max}}$  (see 9.3), is required for this measurement method.

- b) Operate the transmitter without modulation at rated power output. Connect P1 (of figure 8c) to Q2 and adjust the attenuator (12) so that the spectrum analyser (3) is operating in its linear region. Record the value of the attenuation of the attenuator (12) in decibels and the level measured on the spectrum analyser in decibels per metre.  
c) Calculate and record the amount of the coupling loss ( $A_c$ ) of the RFCD using the relationship:

$$A_c = P_{\text{max}} - A - B \text{ dB}$$

where

$P_{\text{max}}$  is the maximum effective radiated power (see 9.3);

$A$  is the attenuation of the attenuator (12) recorded in step b);

$B$  is the level measured by the spectrum analyser (3) in step b).

- d) With switch (4) in position A, connect P1 to P2 and Q1 to Q2. Adjust the attenuator (12) so that the spectrum analyser is operating in its linear range. Record the level measured on the spectrum analyser (3) in decibels per metre.
- e) Adjust the frequency of the radio-frequency test signal source (9) to a frequency 100 kHz above the operating frequency of the transmitter.
- f) Change the switch (4) to position B. Adjust the level of the radio-frequency test signal source to produce a level displayed on the spectrum analyser (3) which is below the level recorded in step d) by an amount:

$$30 - 2 A_c \text{ dB}$$

where  $A_c$  is the coupling loss calculated in step c).

Record the level indicated.

NOTE 2 – This procedure will ensure that the ratio of the levels at the transmitter antenna is 30 dB.

- g) With switch (4) in position A, observe the level of the third-order intermodulation product on the spectrum analyser (3) at the frequency which is 100 kHz below the transmitter operating frequency.
- h) Adjust the line stretcher (5) until the maximum level of the intermodulation product is obtained. Record the level in decibels per metre.
- i) Change switch (4) to position B and note the reading of the intermodulation product. A level less than 10 dB lower than that measured in step h) indicates a source of intermodulation other than the transmitter under test, for example the radio-frequency test signal source (9) or the circulator (7). This shall be eliminated and the test restarted.
- j) Calculate the inter-transmitter intermodulation using the following relationship:

$$ITIM = D - E - 2 A_c \text{ dB}$$

where

$D$  is the spectrum analyzer level recorded in step d);

$E$  is the spectrum analyzer level recorded in step h);

$A_c$  is the RFCD coupling loss calculated in step c).

- k) Calculate the transmitter intermodulation conversion loss (TIMCL) using the following relationship:

$$TIMCL = F - E - 2 A_c \text{ dB}$$

where

$F$  is the spectrum analyser level recorded in step f);

$E$  is the spectrum analyser level recorded in step h);

$A_c$  is the RFCD coupling loss calculated in step c).

NOTE 3 – This method of measurement may be used for other frequency spacings and levels of radio-frequency test signal source (9).

NOTE 4 – The spectrum analyser should have a selectivity sufficiently high so as not to be influenced by the transmitter carrier level when measuring the intermodulation product.

### 11.3.2 Method of measurement using a resistive or a capacitive coupler

- a) Connect the equipment as shown in figure 8d.

NOTE 1 – In order to reduce mismatch error, it is important that the attenuator (2) is coupled to the RFCD with the shortest possible connections.

NOTE 2 – The value of the maximum effective radiated power,  $P_{\max}$  (see 9.3), is required for this test.

- b) Operate the transmitter without modulation at the rated power output. Connect A1 (figure 8d) to B2 and measure the output of the transmitter on the spectrum analyser (8). Record the spectrum analyser level value in decibels per metre.

- c) Calculate and record the amount of the coupling loss ( $A_c$ ) of the RFCD using the relationship:

$$A_c = P_{\max} - A_2 - 20 \text{ dB}$$

where

$P_{\max}$  is the maximum effective radiated power (see 9.3);

$A_2$  is the attenuation of the attenuator (2) in dB.

NOTE 3 – Where the coupling loss of the RFCD is small, the attenuator (2) can be a 10 dB attenuator. For high values of coupling loss, the attenuator (2) should be reduced in value.

- d) Adjust the frequency of the radio-frequency test signal source to a frequency 100 kHz above the operating frequency of the transmitter.
- e) Connect A1 to A2 and B1 to B2. Adjust the level of the radio-frequency test signal source until the value monitored on the spectrum analyser (8) is below the level of the signal from the transmitter by an amount:

$$30 - 2 A_c - 2 A_2 \text{ dB}$$

where

$A_c$  is the coupling loss calculated in step c);

$A_2$  is the attenuation of the attenuator in dB.

NOTE 4 – This procedure will ensure that the ratio of the levels at the antenna is 30 dB.

- f) Measure the intermodulation component by direct observation of the spectrum analyser (8) and record the ratio, in decibels, of the level of the interfering signal to the level of the largest intermodulation component.
- g) Calculate the inter-transmitter intermodulation (ITIM) using the following relationship:

$$\text{ITIM} = X - 2 A_c - 2 A_2 \text{ dB}$$

where

$X$  is the ratio recorded in step f);

$A_c$  is the coupling loss calculated in step c);

$A_2$  is the attenuation of the attenuator (2) in dB.

- h) Calculate the transmitter intermodulation conversion loss (TIMCL) using the following relationship:

$$\text{TIMCL} = X - 2 A_c - 2 A_2 - 30 \text{ dB}$$

where

$X$  is the ratio recorded in step f);

$A_c$  is the coupling loss calculated in step c);

$A_2$  is the attenuation of attenuator (2) in dB.

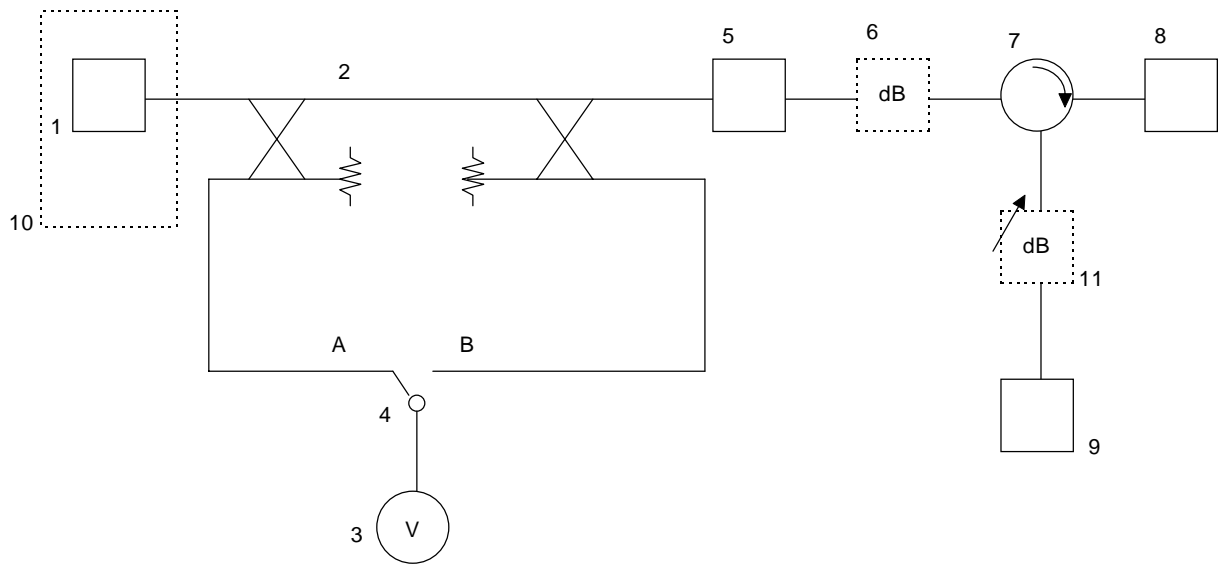
NOTE 5 – The value of intermodulation calculated in step g) is referred to the radiated output of the integrated antenna (ERP) and can be higher than the corresponding value at the antenna terminals depending on the antenna loss. To obtain a result comparable to a transmitter with terminals the antenna loss should be added to the calculated coupling loss in step c).

NOTE 6 – This method of measurement may be used for other frequency spacings and levels of radio-frequency test signal source (9).

NOTE 7 – The spectrum analyser should have a selectivity sufficiently high so as not to be influenced by the transmitter carrier level when measuring the intermodulation product.

Figure 8 – Inter-transmitter intermodulation measuring arrangement

Replace the existing figure 8 with the new figure 8, which is composed of figures 8a, 8b, 8c, and 8d.



IEC 705/99

Key

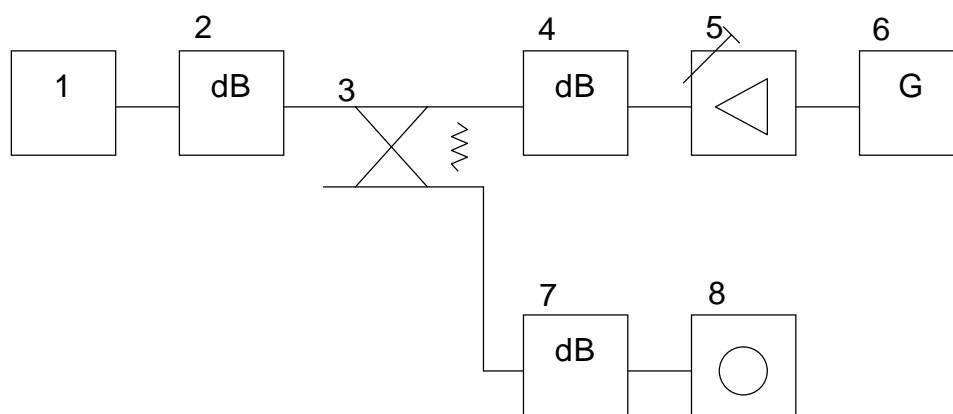
- |   |   |    |                                    |
|---|---|----|------------------------------------|
| 1 | Transmitter under test                                  | 7  | Circulator                         |
| 2 | Two calibrated directional couplers (see notes 1 and 2) | 8  | Test load                          |
| 3 | Selective voltmeter                                     | 9  | Radio-frequency test signal source |
| 4 | Switch  | 10 | Faraday cage (if needed)           |
| 5 | Line stretcher (see note 3)                             | 11 | Attenuator (if needed)             |
| 6 | Attenuator (if needed)                                  |    |                                    |

NOTE 1 –  $M$  dB is the coupling loss, typically 20 dB, and  $N$  dB is the directivity, typically 40 dB.

NOTE 2 – A dual directional coupler can be used in lieu of the two directional couplers. Alternatively, it is possible to use a single directional coupler rotated to measure power from either direction, thereby making the switch (4) unnecessary.

NOTE 3 – A line stretcher is used to maximize the intermodulation distortion.

Figure 8a – Inter-transmitter intermodulation measuring arrangement using a line stretcher and a circulator

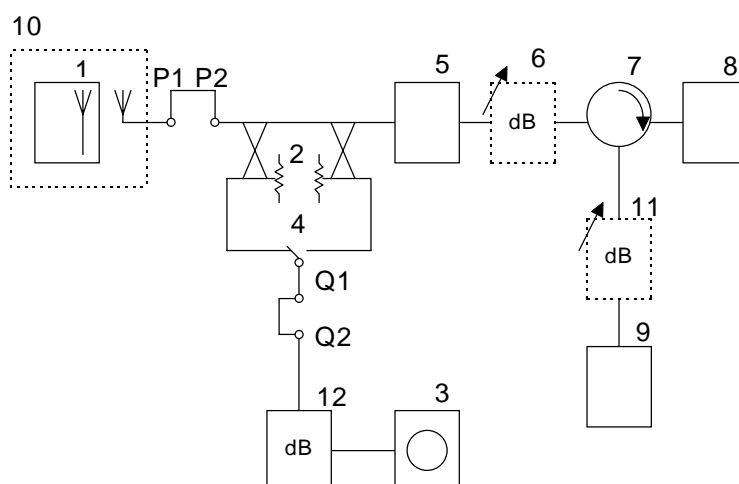


IEC 706/99

### Key

- |   |  |   |                                    |
|---|--|---|------------------------------------|
| 1 | Transmitter under test   | 5 | Power amplifier                    |
| 2 | 10 dB attenuator   | 6 | Radio-frequency test signal source |
| 3 | Resistive coupler (e.g. 1 k $\Omega$ ) or capacitive coupler (e.g. 1 pF) | 7 | Attenuator (if required)           |
| 4 | 20 dB attenuator   | 8 | Spectrum analyser                  |

**Figure 8b – Inter-transmitter intermodulation measuring arrangement using a resistive or a capacitive coupler (antenna terminals)**



IEC 707/99

### Key

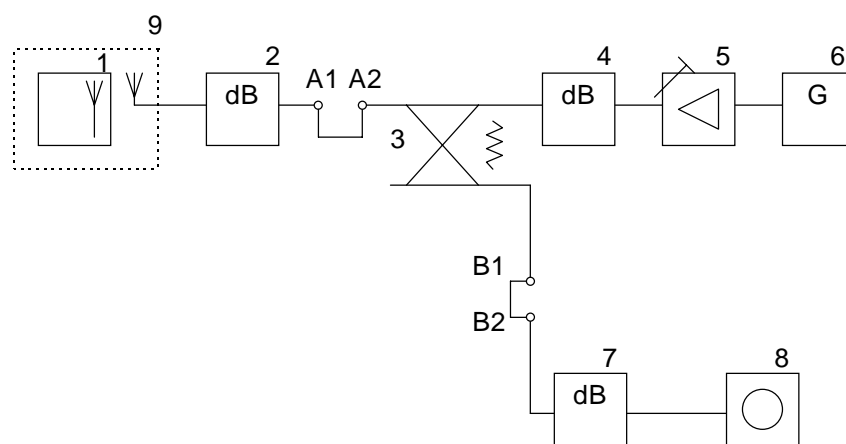
- |   |   |    |                                    |
|---|---|----|------------------------------------|
| 1 | Transmitter under test                                  | 7  | Circulator                         |
| 2 | Two calibrated directional couplers (see notes 1 and 2) | 8  | Test load                          |
| 3 | Selective measuring device (e.g. spectrum analyser)     | 9  | Radio-frequency test signal source |
| 4 | Switch  | 10 | RFCD                               |
| 5 | Line stretcher (see note 3)                             | 11 | Attenuator (if needed)             |
| 6 | Attenuator (if needed)                                  | 12 | Attenuator                         |

NOTE 1 –  $M$  dB is the coupling loss, typically 20 dB, and  $N$  dB is the directivity, typically 40 dB.

NOTE 2 – A dual directional coupler may be used in lieu of the two directional couplers. Alternatively, it is possible to use a single directional coupler rotated to measure power from either direction, thereby making the switch (4) unnecessary.

NOTE 3 – A line stretcher is used to maximize the intermodulation distortion.

**Figure 8c – Inter-transmitter intermodulation measuring arrangement using a line stretcher and a circulator (integral antenna)**



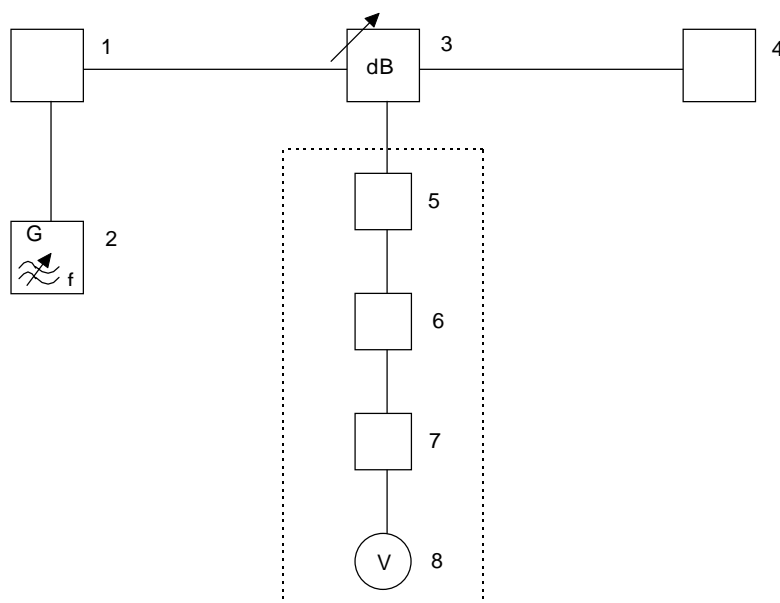
## Key

- |   |  |   |                                    |
|---|--|---|------------------------------------|
| 1 | Transmitter under test   | 6 | Radio-frequency test signal source |
| 2 | Attenuator   | 7 | 20 dB attenuator                   |
| 3 | Resistive coupler (e.g. 1 k $\Omega$ ) or capacitive coupler (e.g. 1 pF) | 8 | Spectrum analyser                  |
| 4 | 20 dB attenuator   |   |                                    |
| 5 | Power amplifier  |   |                                    |
|   |  | 9 | RFCD                               |

IEC 708/99

**Figure 8d – Inter-transmitter intermodulation measuring arrangement using a resistive or a capacitive coupler (integral antenna)**

Replace the existing figure 9 by the following new figure 9:



IEC 709/99

Key

1	Transmitter under test	5	Modulation monitor
2	Audio-frequency generator	6	Band rejection filter (use only in 14.2)
3	Coupler/attenuator	7	De-emphasis network and band-limiting filter
4	Test load	8	Distortion factor meter

NOTE 1 – There is ambiguity in this method of measurement depending upon whether de-emphasis and/or limiting is used in the modulation system. For example

- if the maximum permissible frequency deviation is 5 kHz for any modulating frequency, then 30 % is 1,5 kHz and the audio-frequency signal level that produces 1,5 kHz at 1 000 Hz is the reference signal level;
- if the maximum permissible frequency deviation is 5 kHz at a modulating frequency of 3 000 Hz, and if de-emphasis and/or limiting is not employed, the resulting deviation at a modulating frequency of 1 000 Hz is

$$5 \text{ kHz} \times \frac{1\,000}{3\,000} = 1,66 \text{ kHz}$$

then 30 % of the deviation for a modulating frequency of 1 000 Hz is 500 Hz.

The intention of this method of measurement is to prevent limiting or inadequate signal-to-noise ratios over the audio-frequency band to be measured. Either of the two methods cited may be useful, but alternative a) is preferred.

NOTE 2 – The band rejection filter (6) shall have the following characteristics:

- the relative attenuation of the tone fundamental component shall be at least 40 dB;
- the relative attenuation above twice the tone frequency shall not exceed 0,6 dB;
- in the presence of a noise signal, the filter shall not cause more than 1 dB of reactive attenuation of the total noise output power.

**Figure 9 – Measuring arrangement for modulation**



## 12.4 Presentation of results

*Replace the second paragraph by the following:*

Calculate the modulation characteristic deviations from the reference modulation characteristic, in decibels, taking the deviation at 1 000 Hz equal to 0 dB. The deviation from the reference modulation characteristic, having pre-emphasis of +6 dB/octave, shall be calculated according to the data listed in table 1, unless otherwise specified in the equipment specification.

**Table 1 – Reference modulation characteristic**

Modulation frequency Hz	300	500	1 000	2 000	3 000	3 400
Reference value dB	–10,5	–6,0	0	6,0	9,5	10,6

If pre-emphasis is not provided for in the transmitter, the flat modulation characteristic is considered as reference one in the specified audio-frequency bandwidth.

Page 65

*Insert the following new clause 13:*

## 13 Modulation sensitivity

### 13.1 Definition

Modulation input signal level at which the standard test modulation is provided. It is also called the standard input-signal voltage (see 5.1.4).

### 13.2 Method of measurement

- Connect the equipment as illustrated in figure 9 taking into account 5.1.1.
- Adjust the audio-frequency signal generator (2) to produce the standard test modulation.
- Record the value of the input-signal voltage.

NOTE – The input-signal arrangement used should be recorded.

## 13 Total distortion factor (of a sine-wave at audio-frequencies)

*Replace the title and text of the existing clause 13 as follows:*

## 14 Total distortion factor

### 14.1 Total distortion factor (of a sine-wave at audio frequencies)

#### 14.1.1 Definition

Ratio, expressed as a percentage, of the r.m.s. value of a distorted sinusoidal signal without its fundamental component, to the r.m.s. value of the complete signal. The distorted sinusoidal signal includes harmonically related components, supply ripple and non-harmonically related components.

### 14.1.2 Method of measurement

- a) Connect the equipment as illustrated in figure 9.
- b) Adjust the audio-frequency generator (2) to produce the standard test modulation.
- c) Measure the total distortion factor with a distortion factor meter (8) connected to the output of a modulation monitor (5) or a separate demodulator. The equipment used for measuring the distortion shall be preceded by a network having the appropriate de-emphasis characteristic and by the band-limiting filter described in appendix A.

If required, the measurement may be repeated with other values for the modulating frequency, with the same or other constant values for the resultant modulation level.

## 14.2 Total distortion factor (of two sine-wave at audio frequencies)

This measurement is applicable to a transmitter with a continuous tone control squelch system (CTCSS) in which it is hard to cut out the continuous tone while the transmitter is being measured.

### 14.2.1 Definition

Ratio, expressed as a percentage, of the r.m.s. value of a distorted sinusoidal signal without its fundamental component to the r.m.s. value of the complete signal. The distorted sinusoidal signal includes harmonically related components, intermodulation components, supply ripple and non-harmonically related components.

### 14.2.2 Method of measurement

- a) Connect the equipment as illustrated in figure 9.
- b) Adjust the band rejection filter (6) to obtain maximum attenuation for the tone fundamental component.
- c) Adjust the audio-frequency generator (2) to produce the standard test modulation.
- d) Measure the total distortion factor with a distortion meter (8) connected to the output of a band rejection filter (6). The equipment used for measuring the distortion shall be preceded by a network having the appropriate de-emphasis characteristic and by the band-limiting filter described in appendix A.

If required, the measurement may be repeated with other values for the modulating frequency, with the same or other constant values for the resultant modulation level.

## 14 Relative audio-frequency intermodulation product level

*Renumber clause 14 as clause 15.*

Page 67

*Insert the following new clause 16 before the current clause 15 which should be renumbered as clause 17:*

## 16 Compressor characteristics

The measurement is applicable to transmitters intended for the transmission of angle-modulated signals.

### 16.1 Compressor overall amplitude characteristics

#### 16.1.1 Definition

Relationship between the level of the modulation input, at one frequency, and the resulting deviation of the transmitted carrier.

### 16.1.2 Method of measurement

- a) Connect the test tone to a modulation input.
- b) Connect the transmitter to a deviation meter through an r.f. attenuator.
- c) Set the test tone frequency to 1 kHz.
- d) Operate the transmitter.
- e) Adjust the modulation input level until a deviation of 25 % of maximum permissible frequency or phase deviation is obtained (this input level is the reference level, 0 dB).
- f) Change the level of modulation input which is specified by users and manufacturers, and measure the deviation.
- g) Calculate the relative values of deviation for each level of modulation input.

NOTE – This measurement may be repeated, if desired, at other test tone frequencies.

## 16.2 Compressor attack and recovery time

### 16.2.1 Definitions

Compressor attack time is the time between the instant when a step increase of modulation input level is applied and the instant when the deviation of the carrier falls to a value equal to 1,5 times the new steady-state value.

The compressor recovery time is the time between the instant when a step decrease of the modulation input level is applied and the instant when the deviation of the carrier rises to a value equal to 0,75 times the new steady-state value.

### 16.2.2 Method of measurement

- a) Set the test tone frequency to 2 kHz.
- b) Set the modulation input level so that 25 % of maximum permissible deviation is obtained (this input level is the reference level, 0 dB).
- c) Set the modulation input level to –12 dB.
- d) Switch the modulation input level from –24 dB to –12 dB within 0,5 ms and measure the time taken for the frequency deviation to fall to 1,5 times the steady-state value settled in b) above. Record this time as the attack time.
- e) Switch the modulation input level from –12 dB to –24 dB within 0,5 ms and measure the time taken for the frequency deviation to rise to 0,7 times the steady-state value recorded in c). Record this time as the recovery time.

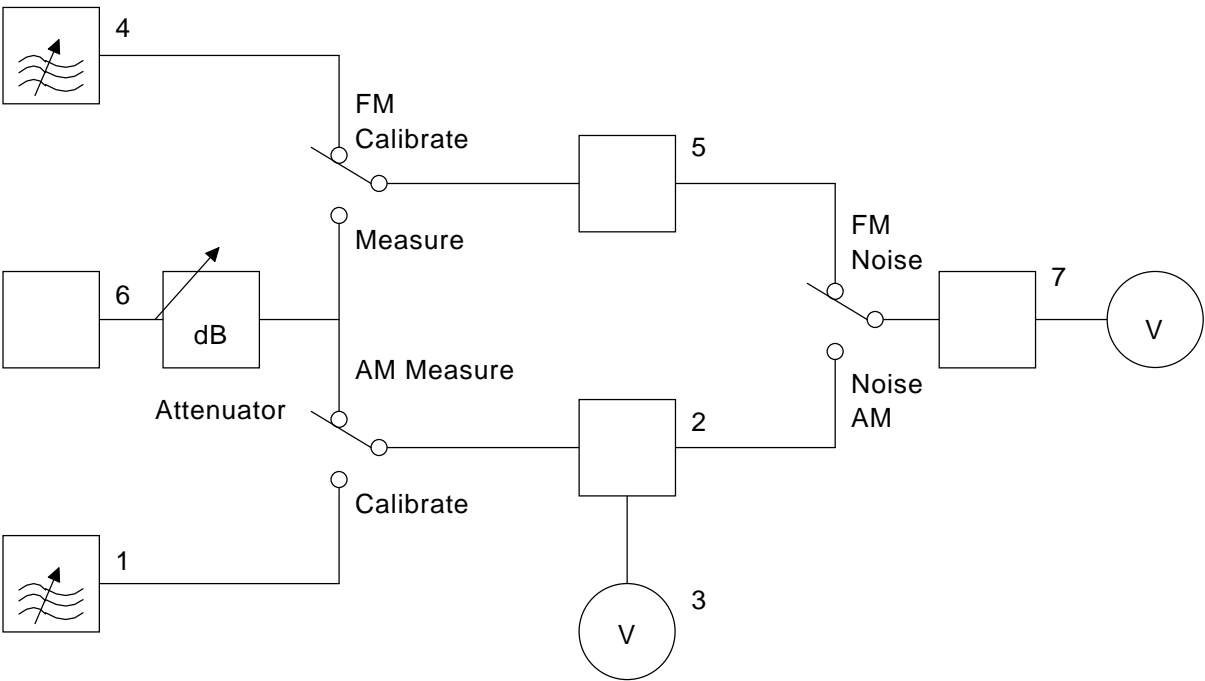
Page 69

## 16 Residual modulation (due to hum and noise)

*Renumber clause 16 as clause 18.*

Figure 11

Replace this figure by the following:



Key

- |   |                                      |   |  |
|---|--------------------------------------|---|--|
| 1 | Amplitude-modulated signal generator | 5 | Frequency-modulation monitor                 |
| 2 | Amplitude-modulation monitor         | 6 | Transmitter under test                       |
| 3 | DC component meter                   | 7 | De-emphasis network and band-limiting filter |
| 4 | Frequency-modulated signal generator | 8 | True r.m.s. voltmeter (a.c. coupled)         |

NOTE – The de-emphasis network in (7) should be used only if pre-emphasis is used in the transmitter under test.

Figure 11 – Measuring arrangement for residual modulation

17 Transmitter attack time

Replace the number and the text of this clause as follows:

19 Transmitter attack time

19.1 Definition

Transmitter carrier switching time or transmitter frequency switching time whichever is longer (see 19.2 and 19.3).

## 19.2 Transmitter carrier switching time

### 19.2.1 Definition

Elapsed time from

- a) the instant of changing the state of the transmitter from standby to transmit or from a frequency channel to another frequency channel; and
- b) the instant at which the unmodulated carrier power reaches a value which is 3 dB below the steady-state value.

### 19.2.2 Method of measurement

- a) Connect the equipment as illustrated in figure 12. The trace of the oscilloscope shall be calibrated in power (Y axis) and in time (X axis), using the signal generator.
- b) Simultaneously, initiate the changing of the state of the transmitter and trigger the horizontal scan of the oscilloscope. Initiation of the transmit function may be carried out by a voice-operated device.
- c) Measure the transmitter carrier switching time by direct reading on the oscilloscope.

## 19.3 Transmitter frequency switching time

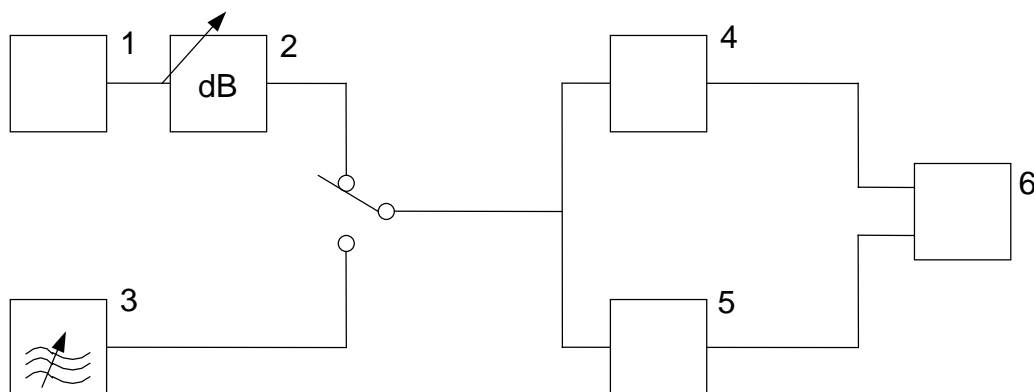
### 19.3.1 Definition

The transmitter frequency switching time is the elapsed time from

- a) the instant of changing the state of the transmitter from standby to transmit or from a frequency channel to another frequency channel; and
- b) the instant after which the frequency of the unmodulated carrier always remains within  $\pm 1$  kHz, unless otherwise specified, of its steady-state frequency.

### 19.3.2 Method of measurement

- a) Connect the equipment as illustrated in figure 12. The trace of the oscilloscope shall be calibrated in frequency (Y axis) and in time (X axis), using the signal generator.
- b) Simultaneously, initiate the changing of the state of the transmitter and trigger the horizontal scan of the oscilloscope. Initiation of the transmit function may be carried out by a voice-operated device.
- c) Measure the transmitter frequency switching time by direct reading on the oscilloscope.



IEC 711/99

Key

1	Transmitter under test	4	RF detector or spectrum analyser
2	Coupler/attenuator	5	Test discriminator
3	Radio-frequency signal generator	6	Storage oscilloscope

NOTE 1 – Subclause 19.2 uses a radio-frequency detector or spectrum analyser (4) and 19.3 uses a test discriminator (5). A dual trace storage oscilloscope can record both simultaneously.

NOTE 2 – The attenuation of the coupler or attenuator (2) shall be chosen in such a way that the input of the test discriminator (5) is protected against overload and the limiter amplifier of the test discriminator operates correctly in the limiting range as soon as the transmitter carrier power (before attenuation) exceeds a level near 1 mW.

**Figure 12 – Measuring arrangement for transmitter attack time**

## 18 Transmitter performance under conditions deviating from standard text conditions

*Renumber the title and text of clause 18 as clause 20.*

Page 77

## Appendix A

### Recommended characteristics of the measuring equipment

*Replace clause A.3 by the following:*

#### A.3 Distortion-factor meter or SINAD meter

The distortion-factor meter or SINAD meter shall have the following characteristics:

- filter attenuation at 50 (60) Hz shall be sufficient, for example 30 dB. The attenuation slope of the filter above the upper cut-off frequency shall be at least 12 dB/octave;
- the band elimination filter attenuation shall be at least 40 dB at 1 000 Hz, but not more than 0,5 dB between 50 Hz and 500 Hz, and between 2 000 Hz and 20 000 Hz.

If the SINAD meter has a fixed filter, it shall have 40 dB attenuation over the frequency band of 980 Hz to 1 020 Hz;

- the noise signal level from a constant amplitude noise source between 300 Hz and 3 000 Hz shall not be attenuated by more than 1 dB by the filter;

- the indicator shall be a true r.m.s. type for a crest factor of 3 or less.

NOTE 1 – SINAD is defined as  $\frac{S+N+D}{N+D}$

NOTE – As a result of the different characteristics of the band elimination filters, the measurement results obtained with the distortion-factor meter or SINAD meter specified above may differ from those obtained with the measuring equipment specified in the previous edition of this standard.

*Replace clause A.5 by the following:*

## A.5 Audio-frequency band-limiting filter

Unless otherwise specified, the upper cut-off frequency of the audio-frequency band-limiting filter shall be about three times the upper frequency limit of the band of modulation frequencies specified for the transmitter. The lower cut-off frequency shall be the lowest modulation frequency specified, with an allowable attenuation of 3 dB at the cut-off frequency.

To be effective, the attenuation slope of the filter shall be at least 12 dB/octave.

In the case of speech transmission, this filter may be replaced by the psophometric filter described in ITU-T Recommendation O.41 (see table below).

**Table A.2 – Characteristics of the psophometric filter**

Frequency Hz	Relative weighting dB	Limit dB
16,66	–85,0	–
50	–63,0	2
100	–41,0	2
200	–21,0	2
300	–10,6	1
400	–6,3	1
500	–3,6	1
600	–2,0	1
700	–0,9	1
800	0	0 (reference)
900	+0,6	1
1 000	+1,0	1
1 200	0,0	1
1 400	–0,9	1
1 600	–1,7	1
1 800	–2,4	1
2 000	–3,0	1
2 500	–4,2	1
3 000	–5,6	1
3 500	–8,5	2
4 000	–15,0	3
4 500	–25,0	3
5 000	–36,0	3
6 000	–43,0	–

Page 79

*Replace clause A.6 by the following:*

## **A.6 Standard test receiver**

The standard test receiver shall be designed to receive signals of the emission class produced by the transmitter under test. Its selectivity, Type B1, shall be measured according to IEC 60489-3, except that the unwanted signal shall be unmodulated.

The minimum selectivity slope, between the points where the ratios of the unwanted to the wanted signals are 6 dB and 85 dB, shall be determined by the channel spacing as follows:

Channel spacings	Minimum selectivity slope
≤15 kHz	12 dB/kHz
>15 kHz	6 dB/kHz

The standard test receiver selectivity measured in conformity with IEC 60489-3, clause 17, shall be more than the value calculated from:

$$S = M + 10 \text{ dB}$$

where

$M$  is the ratio of the transmitter carrier power to the spurious transmitter noise which falls in the specified receiver bandwidth in dB.

The bandwidth at the unwanted-to-wanted input signal ratio of 6 dB shall be indicated.

Page 87

*Delete appendix B.*

Page 93

## **Appendix C**

### **Example of a mains power line impedance stabilization network**

*Rename appendix C as appendix B.*

Page 99

*Delete appendix D.*





ISBN 2-8318-4812-1



---

**ICS 33.060.20; 33.060.30**

---