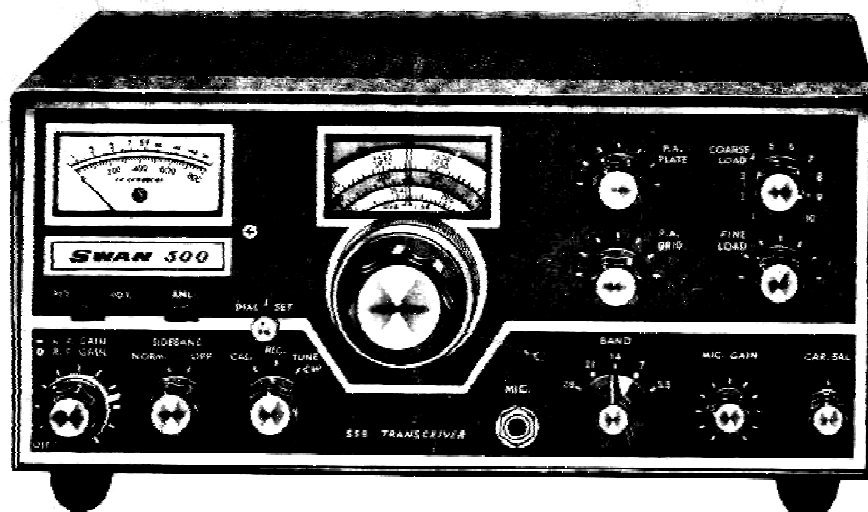




## OPERATION and MAINTENANCE MODEL 500 SERIES Single Sideband Transceiver



### INTRODUCTION

The Swan Model 500 Single Sideband Transceiver together with its accessories and optional equipment, is designed to be used in either CW or SSB modes on all portions of the 80-, 40-, 20-, 15-, and 10 meter amateur radio bands. MARS frequencies may also be covered by using the Model 405X oscillator accessory.

The model 500 generates a single sideband signal by means of crystal lattice filter, and the transceive operation automatically tunes the transmitter to the received frequency. Provisions are included in the transceiver for operation on either upper or lower sideband.

Basic circuitry of the single conversion design has been proven in several thousand of the popular Swan transceivers. Mechanical, electrical, and thermal stability is exceptionally high. All oscillators are temperature compensated and voltage regulated. Push-to-talk operation is standard, with provision for plugging in the Model VX-1 accessory Vox unit for automatic voice control.

The basic transceiver provides coverage of all portions of the 80 through 10- meter amateur bands. In addition to this, the amplifier circuits will tune to most MARS frequencies near the 80-, 40-, and 20-meter bands. By using the Model 405X crystal oscillator accessory, MARS operation is thus possible.

With a suitable power supply, operation may be fixed, portable, or mobile. Power input on all bands exceeds 480 watts, PEP, on single sideband, 360 watts on CW, and 125 watts on AM. The model 500 includes automatic gain control, (AGC), automatic level control, (ALC), automatic noise limiter, (ANL), and grid block keying.

Part I of the instruction manual covers the basic transceiver. Part II covers the recommended power supplies, Model 117-XC for ac operation and Model 14-117 for 12-volt dc operation. Part III provides information on various accessories.

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## SPECIFICATIONS:

### FREQUENCY RANGES

80 Meters	3.5 to 4.0 mc.
40 Meters	7.0 to 7.5 mc.
20 Meters	13.85 to 14.350 mc.
15 Meters	21.0 to 21.5 mc.
10 Meters	28.0 to 29.7 mc.

### POWER INPUT

Single Sideband, Suppressed Carrier:  
480 watts, PEP, minimum on all bands.  
CW:  
360 watts, dc input on all bands.  
AM (Single Sideband with Carrier):  
125 watts dc input on all bands.

### DISTORTION

Distortion products down approx. 30 db.

### UNWANTED SIDEBAND SUPPRESSION

Unwanted sideband down more than 50 db.

### CARRIER SUPPRESSION

Carrier suppression greater than 60 db.

### RECEIVER SENSITIVITY

Less than 0.5 microvolt at 50 ohms impedance  
for signal-plus-noise ratio of 10 db.

### AUDIO OUTPUT AND RESPONSE

Audio output, 4 watts to 3.2 ohm load. Response essentially flat from 300 to 3000 cps in both receive and transmit.

### TRANSMITTER OUTPUT

Wide range Pi-network output matches antennas essentially resistive from 15 to 500 ohms impedance, with coarse and fine load adjustment.

### METERING

Power amplifier cathode current 0-800 ma on transmit, S-Meter 0-70 db over S9 on receive.

### FRONT PANEL CONTROLS

CAL-Rec Tune, CW, AF-RF Gain, Mic. Gain, Bandswitch, Carrier Balance, PA Plate Tune, PA Grid Tune, PA Load Course, PA Load Fine, VOX-PTT Switch, AND Switch, Dial set.

### REAR PANEL CONTROLS AND CONNECTIONS

Bias potentiometer, CW key jack, Jones plug power connector, Vox connector, Antenna jack, S-Meter zero, Auxilliary relay switching, Out-board VFO connector.

### VACUUM TUBE COMPLEMENT

V1	6EW6 VFO Amplifier
V2	12BE6 Transmitter Mixer
V3	6GK6 Driver
V4	6HF5 Power Amplifier
V5	6HF5 Power Amplifier
V6	12BZ6 Receiver RF Amplifier
V7	12BE6 Receiver Mixer
V8	6EW6 First IF Amplifier
V9	12BA6 Second IF Amplifier
V10	12AX7 Product Detector/Receive Audio
V11	6BN8 AGC Amplifier/Rectifier
V12	6GK6 Audio Amplifier
V13	7360 Balanced Modulator
V14	12BA6 Carrier Oscillator
V15	12AX7 Mic. Amplifier/Transmit Audio
V16	0A2 Voltage Regulator
V17	12BA6 100KC Calibrator

### DIODE AND TRANSISTOR COMPLEMENT

Q1	2N706 Oscillator
Q2	2N706 Emitter Follower
D401	TS-2 ALC Diode
D402	TS-2 ALC Diode
D1601	1N2974A Zener voltage regulator
D1602	TS-2 Relay Silencing Diode
D1603	TS-2 AGC Charging bypass

### POWER REQUIREMENTS

Filaments	12.6 volts, 5.5 amps, ac or dc
Relay	12 volts dc, 250 ma.
Bias	-110 volts dc, 100 ma.
Med. voltage	275 volts dc, 150 ma.
High voltage	800 volts dc, 500 ma. Peak Trans.

### DIMENSIONS AND WEIGHT

Height	5-1/2 in.	Depth	11 in.
Width	13 in.	Weight	17-1/4 lb.

## A. CIRCUIT THEORY

## GENERAL DISCUSSION

The Swan 500 transceiver provides single sideband, suppressed carrier transceive operation, and generates the single sideband signal by means of a crystal lattice filter. To permit a logical discussion of this mode of operation, certain definitions are necessary. In a normal AM signal, (double sideband with carrier), a radio frequency signal is modulated with an audio frequency signal. This is considered by many to be merely a case of varying the amplitude of the carrier at an audio rate. In fact, however, there are actually sideband frequencies generated, which are the results of mixing the RF and the AF signals. These sidebands are the sum of, and the difference between the two heterodyned signals. In the detection of this conventional AM signal, the two sidebands are mixed with the carrier to recover and reproduce the audio intelligence. This is an inefficient means of transmission, because only 25 percent of the transmitted power is used to transmit intelligence. There are other attendant drawbacks, also. The bandwidth of AM voice transmission is approximately 6 kc, while the actual demodulated audio is only approximately 3 kc. The result is inefficient use of the frequency band, and over half of the allotted band is unusable due to heterodynes, interference, and congestion.

In the single sideband, suppressed carrier mode of transmission, only one of the sideband signals is transmitted. The other sideband and the carrier are suppressed to negligible level. In addition to increasing the transmission efficiency by a factor of four, single sideband effectively doubles the number of stations or channels which can be used in a given band of frequencies.

It should be remembered that in the single sideband, suppressed carrier mode of transmitting, the unwanted sideband and carrier are only suppressed, not entirely eliminated. Thus, with a transmitted signal from a transmitter with 50 db sideband suppression, the other or unwanted sideband will be present, and will be transmitted, but its level will be 50 db below the wanted sideband. When this signal is received at a level of 20 db over S9, the unwanted sideband will be present at a level of approximately S5. The same is true of carrier suppression. With carrier suppression of 60 db, and a signal level of 20 db over S9, carrier will be present at a level of approximately S3 to S4.

In the Model 500 transceiver, the single sideband suppressed carrier signal is generated by the crystal lattice filter method. For details, refer to the schematic diagram, and to Figures 1, 2, and 3.

## SIGNAL GENERATION

When the push-to-talk switch on the microphone is pressed, the transmitter portion of the transceiver is activated, and it generates a single sideband, sup-

pressed carrier signal in the following manner. Carrier is generated by V14 Carrier Oscillator, which is a Pierce oscillator with the crystal operating in parallel resonance. This stage operates in both the transmit and receive modes. When transmitting, the RF output of the oscillator is injected into the control grid of the Balanced Modulator, V13. This balanced modulator is a beam deflection tube, and operates similar to a cathode ray tube in that the electron beam from the cathode is deflected to one output plate or the other by the charge appearing on the deflection plates. The carrier signal fed to the control grid of the balanced modulator appears on both plates of the output. The two plates are connected to Transformer Z1301 in push-pull, so the carrier signal cancels itself out in Z1301. The deflection plate DC voltages are adjusted by means of the carrier balance control so that the RF being fed to the output plates will cancel out, and the output from Z1301 will be zero. Audio signals from the Microphone Amplifier, V15, are applied as a modulating voltage to one deflection plate, and the two sidebands resulting from the sum and difference frequencies of the audio and carrier signals appear in the output of transformer Z1301. Carrier suppression is approximately 60 db.

The double sideband, suppressed carrier signal is then coupled from the secondary winding of Z1301 to the crystal filter, which suppresses the lower sideband, and permits only the upper sideband to be fed to the First IF Amplifier, V8. The carrier frequency is generated at approximately 5172.8 kc, normal sideband. With the opposite sideband crystal, the carrier crystal frequency will be 5176.3 kc, and this positions the double sideband signal on the other side of the filter response curve, attenuating the upper sideband by at least 50 db. In the single conversion mixing process, these sidebands become inverted on 80 and 40 meters. Thus the Swan-500 normally operates on lower sideband on 80 and 40, while on 20, 15, and 10 meters normal operation is on upper sideband.

Q1, the VFO 2N706 Oscillator, operates in the common base configuration as a Colpitts oscillator. Q2, the Emitter Follower is used for isolation and impedance matching purposes. The extremely good regulation achieved through using the Zener diode regulator D1601 across the bias supply voltage, also contributes to the stability. Bandswitching is accomplished by changing the tank circuit coil. The VFO in the Model 500 exhibits extremely good stability after the initial warm-up period. Drift from a cold start will be less than 1 kc for the first hour on 80-, 40-, and 20-meter bands, and less than 2 kc on 10 and 15 meters. After the initial warm-up period, drift will be negligible.

The single sideband, suppressed carrier signal from the First IF Amplifier is fed to the Transmitter.

# I MODEL 500 TRANSCEIVER

## A. Circuit Theory (Cont)

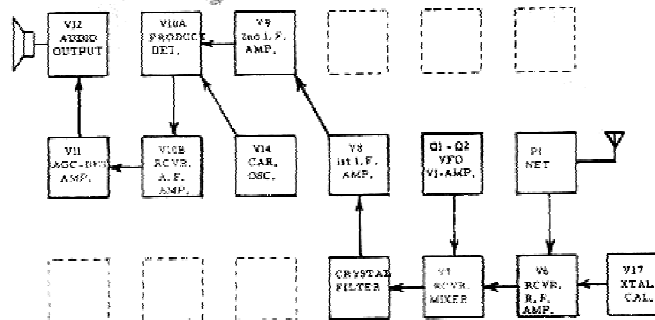


FIGURE 1 BLOCK DIAGRAM, RECEIVE MODE

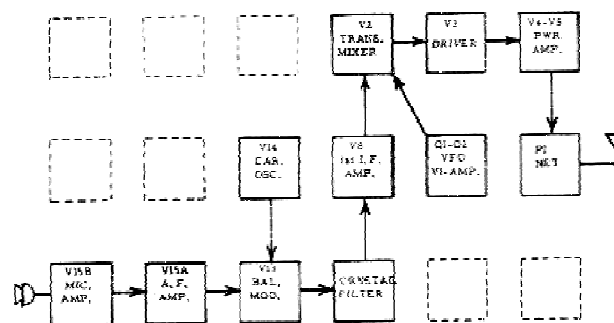


FIGURE 2 BLOCK DIAGRAM, TRANSMIT MODE

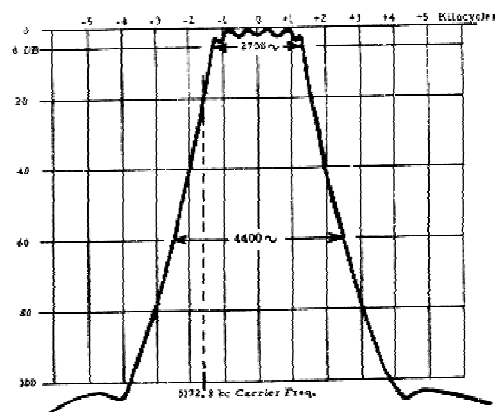


FIGURE 3 CRYSTAL FILTER, TYPICAL CHARACTERISTIC

## MODEL 500 TRANSCEIVER

### A. Circuit Theory (Cont)

#### SIGNAL GENERATION (cont)

Mixer, V2, where it is heterodyned with the VFO signal. The resultant signal at the desired transmit frequency is amplified by the Driver, V3, and the Power Amplifiers, V4 and V5. The signal from the VFO Amplifier is initiated in the transistorized VFO-Emitter Follower circuit Q1, Q2. The signal from the VFO is routed to the VFO Amplifier, and on 40 and 80 meters, is subtractively mixed with the single sideband signal from the IF Amplifier to result in LSB operation. On 20, 15, and 10 meters, the frequencies are additively mixed, resulting in output on the upper sideband.

When in TRANSMIT, the gain of the First IF Amplifier is controlled through the Automatic Level Control network D401-D402, etc., to control the gain of the stage in response to the average input power to the Power Amplifiers. This ALC system will compensate for any extremely strong input signals, but does not completely eliminate the necessity of proper adjustment of the Mic. Gain Control. This feature will help prevent the transmitter from flat topping and spurious emissions, but considerable distortion may occur if the Mic. Gain Control is not properly adjusted. Refer to Operating Instructions.

#### TUNE AND CW OPERATION

(See page 9 for Tuning Instructions)

Normally, the frequency of the carrier oscillator is approximately 300 cps outside the 6 db passband of the crystal lattice filter. In TUNE position, the frequency of the carrier oscillator is moved approximately 500 cps to place it well within the passband of the crystal lattice filter. A similar procedure is followed for CW to allow full carrier output during CW operation. During CW operation, the cathode of V15A is disconnected from ground. This allows CW operation with no accidental audio modulation from the microphone.

#### RECEIVE

In RECEIVE position, or at any time when the transmitter is not in TRANSMIT, all circuits used in transmitting are disabled through the relay controlled circuits, K1, K2. The relays are energized for transmitting and de-energized for receiving. Relay K2, when de-energized, allows signals from the transmitting tank circuit and antenna to be fed to the Receiver RF Amplifier, V6, where they are amplified and then fed to the control grid of the Receiver Mixer, V7. The local oscillator signal from the VFO Amplifier is now used to heterodyne the received signal to the IF frequency. All IF amplification is accomplished at this frequency, nominally 5174.5 kc, through V8 and V9 IF amplifiers. In the Product Detector V10A, the IF signal is heterodyned with the carrier frequency generated by Carrier Oscillator, V14. The resultant audio signal is then put thru a noise clipper network

which is in or out of circuit by selection of ANL switch. The signal is then amplified by V10B, which then couples to V11, the AGC amplifier, and V12, the output audio stage.

#### FREQUENCY CALIBRATION

Frequency calibration of the Model 500 is in 5 kc increments on 80, 40, 20, and 15-meters, and in 20 kc increments on 10M. 80- and 20-meters are calibrated directly on the upper dial scale. 40- and 15-meters are calibrated from zero to 500 on the green tinted center scale, and 10-meters is calibrated directly on the lower dial scale.

The two red indicator lines on the dial window represent the passband of the transceiver, and the actual carrier frequency depends on which sideband is in use. For upper sideband, carrier frequency is read with the left hand indicator line. For lower sideband, use the right hand line.

Dial accuracy and tracking are very good on the 500, but caution must always be observed when operating near band edges. Measuring the frequency with the 100 kc calibrator when working near band edges is recommended.

#### DIAL SET

A dial-set has been provided so that dial adjustment can be made at any 100 kc point of the dial. With calibrator on, set the dial to any 100 kc point closest to the frequency you wish to work. Now adjust dial-set control to zero-beat the VFO with the 100 kc Calibrator. This provides greater accuracy of dial readout.

**CAUTION:** Care must be exercised when tuning for the 100 kc harmonics of the calibrator. Several spurious image signals may be heard, although they will be somewhat weaker than the actual harmonics.

#### TRANSMIT AND RECEIVE SWITCHING

Transmit and receive switching is performed by relays K1 and K2. In TRANSMIT position, only those tubes that operate in the transmit mode are operative, all others being biased to cutoff through the relay contacts. In the RECEIVE position, with the relays de-energized, the tubes that are used only in transmit are cut off in the same manner. Relay K2 when de-energized, feeds signals from the output pi-network to the receiver, and is used also to control external switching circuits. In transmit position the meter indicates the combined cathode current of the two power amplifiers. In receive position, it indicates the voltage across R903 in the screen grid of the Second IF Amplifier, V9, which is inversely proportional to the AGC voltage used to control the gain of the tube. Thus, the meter indicates the relative strength of received signals.

#### POWER RATING

The Swan 500 is capable of 480 watts, PEP input under steady state two-tone test conditions, when operated with any of the recommended power supplies. The peak envelope power, when voice modulated, is considerably greater, typically 500 watts, or more.

## I MODEL 500 TRANSCEIVER

### A. Circuit Theory (Cont)

Recommended power supplies produce a no-load plate voltage of approximately 925 volts. Under TUNE conditions, or CW operation, this voltage will drop to approximately 720 volts. Under steady state two-tone modulation, the voltage will drop to approximately 750 volts. If the power amplifier idling current is 50 ma, and the two-tone current, just before flat-topping, is 375 ma, the peak two-tone current will be 560 ma. Under these conditions the PEP input will be 750 volts times 560 ma, = 420 watts. Under voice modulation, because average power is considerably less, the power amplifier plate and screen voltages will be maintained higher, even during voice peaks, by the power supply filter capacitors. Peak plate current will therefore also be higher than with two-tone test conditions. Under typical operating conditions, peak plate current before flat-topping will be 625 ma at 800 volts, to result in an input of 500 watts, PEP. Readings of cathode current will not reflect this 500 watt power input, however, because of the damping in the cathode current meter. Cathode current readings under normal voice input should not exceed approximately 150 to 175 ma.

### POWER AMPLIFIER PLATE DISSIPATION

There is often a misunderstanding about the plate dissipation of tubes operated as AB1 amplifiers under voice modulation. In the Swan 500, while in the transmit position, and with no modulation, the plate voltage will be 890 volts, the plate current 60 ma, and the power input will be 45 watts.

Authorities agree that the average voice power is 10 to 20 db below peak voice power. Normally some peak clipping in the power amplifier can be tolerated, and a peak-to-average ratio of only 6 db may sometimes occur. Under such conditions, the average power input will be 125 watts, and average plate current will be 156 ma. With power amplifier efficiency of 65 percent, plate dissipation will be 44 watts, or 22 watts per tube. The 6HF5 is rated at 28 watts, continuous duty cycle, in normal TV service. Thus it can be seen that under normal operating conditions, the power amplifier tubes in the Swan 500 are not being driven very hard. Note, however, that proper modulation level must be maintained by correct setting of Mic. Gain, and that the length of time in TUNE position must be limited to not more than 30 sec. at a time.

### B. INSTALLATION

#### GENERAL

The Swan 500 transceiver has been designed to provide the utmost in ease of operation, stability, versatility, and enjoyment. Maximum enjoyment from your Swan will depend to a great extent on the installation. For fixed station or portable use, operation with the Model 117-XC power supply

provides a compact arrangement with maximum ease of operation. All switching is performed in the transceiver. For mobile installations, the Model 14-117 supply provides similar switching arrangements, and speaker output may be fed to an optional 3 x 5 speaker with mounting provisions on the inside of 500 cabinet. Alternately, the speaker output may be fed through the car broadcast radio speaker.

### POWER SUPPLY

The Swan Models 117-XC or 230-XC Power Supplies provide all necessary voltages required by the transceiver for AC operation. The supplies come equipped with a pre-wired plug and cable, all ready for plugging into the transceiver. The Model 14-117 supply for mobile operation includes all necessary cables, connector plug, fuses, and installation hardware. The Jones plug for connection to the transceiver is furnished with the unit.

Power requirements for the Swan 500 are listed in the following table. Pin connections to the Jones type power connector are listed as an aid in connecting other brands or home-brew supplies.

### EXTERNAL SPEAKER CONNECTIONS

Audio output from the transceiver is provided at pin 12 of the Jones plug, and to a terminal lug located near V12 audio output tube on top of chassis for optional "3 x 5" internal speaker. The other speaker lead goes to the common chassis ground at pin 6 of Jones plug or to ground terminal at the terminal strip. For mobile installations, mounting a "3 x 5" internal speaker makes the installation simple. In case you wish to use the car broadcast radio speaker a DPDT selector switch should be installed to select either the broadcast radio or transceiver output. (See Figure 8).

Provisions for headphones will be found on front of AC Power Supply. (See Page 17) It is recommended that high impedance phones be used for best results.

### JONES PLUG CONNECTIONS

	Pin	Nominal	Minimum	Maximum
High Voltage	8	800 VDC 500 MA	600 VDC Low Pwr.	1200 VDC Hi. Power
Medium Voltage	10	275 VDC 150 MA	225 VDC	325 VDC
Bias Voltage	3	-110 VDC 100 MA	-100 VDC	-130 VDC
Filament Voltage	4	12.6 V* 5.5 amp	11.5 V	14.5 V
Relay Voltage	5	12 VDC 250 MA	10 VDC	14.5 VDC
*AC or DC				

## MODEL 500 TRANSCIVER

### B. Installation (Cont)

#### MICROPHONE

The microphone input is designed for high impedance microphones only. The choice of microphone is important for good speech quality, and should be given serious consideration. The crystal lattice filter in the transceiver provides all the restriction necessary on audio response, and further restriction in the microphone is not required. It is more important to have a microphone with a smooth, flat response throughout the speech range. The microphone plug must be a standard 1/4 in. diameter three-contact type. The tip connection is for push-to-talk relay control, the ring connector is the microphone terminal, and the sleeve is the common chassis ground. The microphone manufacturer's instructions should be followed in connecting the microphone cable to the plug. With many microphones, the push-to-talk button must be pressed to make the microphone operative. For VOX operation, this feature may be disabled, if desired, by opening the microphone case and permanently connecting the contacts which control the microphone.

#### ANTENNA

Any of the common antenna systems designed for use on the high frequency amateur bands may be used with the Swan transceiver, provided the input impedance of the transmission line is not outside the capability of the pi-output matching network. The transmission line should be of the coaxial cable type. An antenna system which shows a standing wave ratio of less than 4:1 when using 50 or 75 ohm coaxial transmission line, or a system that results in a transmission line input impedance that is essentially resistive, and between 15 and 500 ohms will take power from the transceiver with little difficulty. If open-wire or balanced type transmission line is used with the antenna, a suitable antenna tuner is recommended between the transceiver and the feedline. Methods of constructing and operating such tuners are described in detail in the ARRL Antenna Handbook, and similar publications. For operation on the 75- and 40- meter bands, a simple dipole antenna, cut to resonate in the most used portion of the band, will perform satisfactorily. For operation on the 10, 15, and 20 meter bands, the efficiency of the station will be greatly increased if a good directional rotary antenna is used. Remember that even the most powerful transmitter is useless without a proper and efficient antenna system.

#### MOBILE ANTENNA

Mobile antenna installations are critical, since any mobile antenna for use on the high frequency bands represents a number of compromises. Many amateurs lose the efficiency of their antenna through improper tuning. Points to remember about the mobile antenna used with the Swan 500 are:

1. The "Q" of the antenna loading coil should be as high as possible. There are several commercial models available which use high "Q" coils, including the Swan Model 45 and Model 55 5 band "Swantennas".
2. The loading coil must be capable of handling the power of the Model 500 without overheating. In TUNE position, the power output of the transceiver may exceed 250 watts. Wide spaced, heavy wire loading coils are essential.
3. The SWR bridge is a useful instrument, but unfortunately it is quite often misunderstood, and over rated in importance. Basically, the SWR bridge will indicate how closely the antenna load impedance matches the transmission line. With long transmission lines, such as will be used in many fixed station installations, it is desirable to keep the impedance match fairly close in order to limit power loss. This is particularly true at the higher frequencies. The longer the line, and the higher the frequency, the more important SWR becomes. However, in mobile installations the transmission line seldom exceeds 20 feet in length, and an SWR of even 4 to 1 adds very little to power loss. The only time SWR will indicate a low figure is when the antenna presents a load close to 50 ohms, but many mobile antennas will have a base impedance as low as 15 or 20 ohms at their resonant frequency. In such a case, SWR will indicate 3 or 4 to 1, and yet the system will be radiating efficiently.
4. The really important factor in your mobile antenna is that it should be carefully tuned to resonance at the desired frequency. The fallacy in using an SWR bridge lies in the fact that it is sometimes possible to reduce the SWR reading by detuning the antenna. Field strength may actually be reduced in an effort to bring SWR down. Since field strength is the primary goal, we recommend a Field Strength Meter for antenna tuning.
5. For antenna adjustments, the Swan-500 may be loaded lightly to about 100 ma. cathode current instead of the usual 500 ma. This will limit tube dissipation during adjustments, and will also help reduce interference on the frequency. In any case, do not leave the transmitter on for very long at one time. Turn it on just long enough to tune and load, and get a field strength reading.

Start out with the antenna whip at about the center of its adjustment range. Set the VFO to the desired operating frequency and then adjust P. A. TUNE for dip, and P. A. LOAD for 100 ma. Then observe the field strength reading. The Field Strength Meter may be set on top of the dash, on the hood, or at an elevated location some distance from the car.

Change the whip length a half inch, or so, at a time, retune the P.A. for 100 ma. loading each time, and check field strength. Continue this procedure until the point of maximum field strength is found. This adjustment

will be most critical on 75 meters, somewhat less critical on 40, etc., until on 10 meters the adjustment will be quite broad. After tuning the antenna to resonance, load the P. A. to full power.

#### CONTROL FUNCTIONS

<b>ON-OFF SWITCH</b> (On AF & RF Gain Knob)	Turns power supply on and off	<b>AF GAIN</b>	Controls potentiometer R1201 in grid circuit of V12 AF Output, and controls audio volume.
<b>CAL-REC-TUNE-CW</b> Calibrate	All voltages are applied to transceiver. Grounds cathode of V17. Removes ground from cathode of V15A.	<b>MAIN TUNING</b>	Controls C1706 in frequency determining tank circuit of VFO.
Receive	All voltages are applied to transceiver.	<b>PA GRID</b>	Controls C1A and C1B in plate tanks of transmitter mixer and driver.
<b>Transmit</b> (Push-to-Talk)	12 volt dc circuit through relay K1 and K2 is completed, and tubes used only in receive are biased to cutoff.	<b>PA TUNE</b>	Controls C417 on pi-network to tune final power amplifier plate to resonance.
Tune-CW	All circuits for transmit are energized, as above. Capacitor C1401 in the carrier oscillator is removed from ground.	<b>PA LOAD, Fine</b>	Controls C420 in pi-network to match impedance of output load. Tunes input to Receiver RF Amplifier.
<b>MIC. GAIN</b>	Controls potentiometer R1503 in the grid of V15A and controls amount of audio to the balanced modulator.	<b>PA LOAD, Coarse</b>	Switches in progressively more capacitance in parallel with PA Load, Fine.
<b>CAL. BALANCE</b>	Controls potentiometer R1305 in the balanced modulator deflection plate circuit, and permits nulling out the carrier.	<b>MAIN BANDSWITCH</b>	Switches, plate coils, and associated capacitors of VFO, VFO Amplifier, V1, Transmitter Mixer, V2, and Driver, V3. Also switches tank coil of pi-coupling system and associated capacitors in PA output tank.
<b>RF GAIN</b>	Controls variable resistor R609, common in the grids of receiver mixer V7, RF Amplifier, V8 and V9 IF Amplifiers.		

#### MODEL 500 TRANSCEIVER

##### C. OPERATION

Before connecting any cables to the Swan 500 perform the following steps:

1. Rotate the PA BIAS control on the rear chassis apron, fully counter clockwise.
2. Rotate the CAL-REC-TUNE-CW on the lower left of the front panel counter clockwise to REC.
3. Rotate the AF GAIN Control counter clockwise to operate the power switch to OFF.

##### POWER SUPPLY, ANTENNA AND GROUND CONNECTIONS

1. Connect wire from earth ground to ground stud provided on rear of chassis.
2. Connect a 50 to 75 ohm antenna feed-line to the coaxial connector on the rear chassis panel.
3. Connect the power supply cable to the Jones connector on the rear chassis apron.

4. Connect the power supply to the proper voltage source.

The Swan Model 500 may be operated from 117 volts, ac, 50 - 60 cycle power with the Model 117-XC power supply, or from 230 volts, 50 - 60 cycles with the Model 230-XC. The Model 500 may be operated from a 12 volt dc source with the Swan Model 14-117 power supply.

##### WARNING

DANGEROUS HIGH VOLTAGE IS PRESENT ON THE PLATE OF THE POWER AMPLIFIER WHENEVER THE POWER SUPPLY IS ENERGIZED. NEVER TURN POWER ON WHEN THE POWER AMPLIFIER COVER IS REMOVED. HIGH VOLTAGE IS ALSO PRESENT AT PIN EIGHT OF THE POWER PLUG.



## RECEIVE OPERATION

1. Rotate the AF GAIN Control clockwise to about the 3 o'clock position. The power switch will operate applying filament, relay, bias, medium, and 800 volt high voltage to the transceiver.
2. Wait approximately one minute to allow the tube filaments to reach operating temperature. During this period, perform the following steps:
  - (a) Rotate the BANDSWITCH to desired band.
  - (b) Rotate MIC. GAIN fully counter-clockwise.
  - (c) Rotate CAR. BAL. control to the mid-scale position, with white dot on knob aligned with the long index mark on the panel.
  - (d) Preset PA PLATE control to mid-position.
  - (e) Preset PA GRID control to mid-position.
  - (f) Preset PA LOAD FINE to mid-position.
  - (g) Preset PA LOAD COARSE to position 1.
  - (h) Set tuning dial to desired operating frequency.
  - (i) Set RF GAIN control to approximately 3 o'clock position.
3. Carefully adjust the PA GRID and the PA PLATE controls for maximum receiver noise. Note: The PA GRID control resonates the transmitter driver stages and the receiver RF amplifier plate circuit. The PA PLATE and PA LOAD controls adjust the input and output capacitors in the transmitter power amplifier final plate circuit, as well as the receiver RF amplifier grid circuit. Proper adjustment of these controls in the receiver position will result in approximately resonant conditions in the transmitter stages.

## RECEIVER TUNING — IMPORTANT, READ CAREFULLY.

Precise tuning of a single sideband signal is very important. Do not be satisfied to merely tune until the voice can be understood, but take the extra care of setting the dial to the exact spot where the voice sounds natural. Above all, avoid the habit of tuning so that the voice is pitched higher than normal. This is an unfortunate habit practiced by quite a number of operators. The following points help to explain the effects of mistuning:

1. If you tune so the received voice is higher than normal pitch, you will then transmit off frequency, and your voice will sound lower than normal pitch to the other station. He will probably retune his dial to make you sound right. If you keep this up, you'll gradually waltz one another across the band. If both of you are mistuning to an unnatural higher pitch, you'll waltz across the band

twice as fast. (And someone will no doubt be accused of frequency drift).

2. Mistuning results in serious harmonic distortion on the voice, and should be quite noticeable to the average ear. Some will claim that if they don't know how the other person's voice actually sounds, they can't tune him in properly, but this is not true. With a little practice, it will be fairly easy to tell. Some voices are relatively rich in harmonics, and are easier to tune in than a person with a "flat" voice. Also, a transmitter which is being operated properly with low distortion will be easier to tune in than one which is being over-driven and is generating excessive distortion. There is no mistaking when you have a station tuned right on the nose. It will sound just like "AM", so to speak. Mainly, avoid the habit of tuning so everyone sounds higher than normal pitch, or like Donald Duck. This is incorrect, unnecessary, and sounds terrible.
3. A vernier control for receive frequency, sometimes referred to as "incremental tuning", is not available on the Swan-500. Such a device is not necessary if proper tuning habits are exercised.
4. Your Swan-500 will automatically transmit on exactly the same frequency as the one to which you are listening. There is no adjustment for making them the same, since by using the same oscillator for both send and receive, it happens automatically. If separation of receive and transmit frequency control is desired, the model 405B or 410 VFO unit may be used. In this case the model 22 dual VFO adaptor must be installed in the 500.

## TRANSMITTER TUNING

### CAUTION

The Model 500 covers several frequency ranges outside the amateur bands. Care must be exercised not to transmit on these frequencies.

Tuning of the transmitter is not complicated, provided the few simple steps are followed in the correct order. Do not attempt initial tuneup without first performing the procedures for Receive operation described above. The following procedures assume that the unit has been checked out in Receive position, and a high impedance push-to-talk microphone is inserted in the MIC. JACK.

1. Press Push-to-Talk to place unit in TRANSMIT, read the cathode current on the meter.
2. Quickly rotate the CAR. BAL. control on the front panel until the meter reads minimum cathode current.
3. Next, adjust the PA BIAS control on the rear of the chassis until the meter reads 50 ma.

4. If this is the first time the transceiver is being tuned on this band, set the PA LOAD switch to position 5. After experience in tuning up, the control may be set to whatever position has been found to be optimum on each respective band. Now, in rapid succession:

- (a) Turn the CAR. BAL. control clockwise until a slight increase in meter reading is obtained.
- (b) Rotate the PA GRID control for maximum meter reading.
- (c) Rotate the PA PLATE control for minimum meter reading.
- (d) Adjust car. bal. for a reading of 150 ma.

**IMPORTANT**—Tuning the PA PLATE for minimum, or "dip", is known as "resonating" the power amplifier plate circuit, and is very important to preserving tube life. If the transceiver is held in Transmit or TUNE position for more than a few seconds while out of resonance and with some grid drive, the 6HF5 tubes may be severely damaged. For this reason we repeat: **CAUTION**—Do not hold the transceiver in Transmit or TUNE position for any length of time without "dipping" the PA PLATE control. The PA GRID must first be "peaked" as in (b), above, and this requires some carrier supplied as described in (a), so it can be seen that these steps must be performed quickly. If the PA LOAD control is too far clockwise, it may not be possible to find a "dip" with the PA PLATE control. For this reason, be sure to observe the first sentence in this section, Step 4.

5. Rotate the CAL-REC. TUNE switch to TUNE position. Quickly check the PA PLATE control for "dip" or minimum reading. If the meter dips to less than 500 ma., increase loading by rotating the PA LOAD controls clockwise. After each increase in PA LOAD resonate the PA PLATE again; that is, adjust it for dip. Continue increasing PA LOAD until the PA PLATE dips to 450-500 ma. Then switch back to RECEIVE. **NOTE:** For 10 meter operation it will be necessary first to repeak the PA grid control for maximum output in the tune position.

**CAUTION:** Do not hold the transceiver in TUNE position for more than 30 seconds at a time, even though PA PLATE is resonated. With full grid drive to the 6HF5 PA tubes, which you have in TUNE position, they are dissipating considerably more power than they do during normal voice transmission, so a short tuning period must be observed.

6. Under some conditions, it may not be possible to load up to 500 ma. This may occur with lower than normal line voltage or tubes not quite up to par, particularly on 10 and 15 meters. The current increase when tuning the

plate circuit off resonance will provide a clue as to how far the power amplifier can be loaded. If the meter swings up to 600 or 700 ma. on either side of resonance, it will be easy to load up to 500 or even more. But, if the tubes draw just 500 ma. off resonance, you can only load to 400 or 450 ma. This is not necessarily a sign that you have a problem. Peak input power with voice modulation will still be 400 watts when you load to 400 ma. in TUNE position. A new pair of PA tubes may allow you to load higher, or possibly a new driver tube will help. Primarily, the level to which you can load will serve as an indication of when tubes are deteriorating. If you can load to 500 ma. when the set is new, and after a few months of operating you cannot get above 400 ma., or so, it is probably time to replace the 6HF5 tubes, and possibly the 6GK5 driver. The other tubes should also be checked at that time.

#### 7. AVERAGE PA LOAD SWITCH POSITIONS.

The following positions are for a 50 ohm non-inductive load, and indicate approximately where the PA LOAD switch will end if the antenna and coaxial cable are well matched.

BAND	PA LOAD SWITCH
80	POS. 7
40	8
20	9
15	9
10	10

A large deviation from these positions indicates a possible matching problem, although operation may still be quite satisfactory. PA LOAD switch positions below 5 will generally be needed only with very low impedance loads, such as a 75 meter mobile antenna with center loading coil.

8. **VOICE TRANSMISSION.** After tuning up as outlined above, press the Push-to-Talk button on the mike and carefully set the CAR. BAL. control for minimum meter reading. While speaking into the mike, slowly rotate the MIC. GAIN control until occasional peak reading of 175 to 200 ma. are obtained. With most microphones, the MIC. GAIN control will be set between 9 and 12 o'clock, but it may vary considerably. The ALC circuit will help limit cathode current to about 200 ma., but turning the MIC. GAIN up too high will still produce flat-topping and spurious signals, so it is important to hold it down. The meter is quite heavily damped, and its reading with average voice modulation may not look very impressive, but the voice peaks are going well over the 480 watt power rating of your Swan transceiver, and signal reports will verify this fact. **NOTE:** Transceiver will not modulate with Function Switch in CAL. position.

8. **TRANSMITTER TUNING WITH SWR BRIDGE OR FIELD STRENGTH METER.** If either of these instruments is available, they are highly recommended as a better method of tuning the PA Amplifier, since they provide a direct indication of relative output. With the SWR Bridge in Forward position, or with the Field Strength Meter set to pick up a portion of the radiated power, simply adjust the PA TUNE and PA LOAD controls for maximum output. This must be done quickly, limited to about 30 seconds, to limit tube dissipation as previously mentioned. This method will result in maximum possible output and efficiency, as well as maximum linearity. You will probably find that cathode current readings end up somewhat less than 500 ma. on 10 meters because grid drive is the least on this band. On 80 meters where grid drive is the greatest, maximum output will be reached at more than 500 ma. These are a normal condition.

**NOTE**—The cathode current level to which the PA is loaded will have no bearing on tube life. When transmitting with normal voice modulation, average power input will be the same regardless of how high or low the PA was loaded while tuning. Peak output, linearity, and lowest distortion will go along with maximum loading. In other words, you will not extend tube life by loading to a lesser degree. The secret to long tube life is simply to keep TUNE-up periods short and not too frequent.

#### AM OPERATION (Single Sideband With Carrier)

1. Tune transmitter to full output on single sideband as described above.
2. Rotate MIC GAIN control to minimum full CCW.
3. With Push-to-Talk pressed, rotate CAR. BAL. control until cathode current is approximately 150 ma.
4. While talking in a normal tone of voice into the microphone, increase MIC GAIN setting until the meter kicks upward slightly. This setting will result in excellent AM transmission.

#### CW OPERATION—SPECIAL NOTE

1. Tune transmitter to full output same as for SSB operation. Then adjust carrier balance for a meter reading of approximately 2/3 the maximum tune-up reading. For example: If tune-up results in 400 ma. cathode current, adjust carrier balance for approximately 275 ma., or if tune-up reading is 500 ma., then reduce to approximately 350 ma. for CW, etc. Note: This reduction in drive is essential on 15 meters in order to eliminate a possible spurious radiation.
2. Insert CW Key in the Key Jack provided on the back of the 500. Use a standard 1/4 inch diameter 2 circuit phone plug.
3. Switch to TUNE-CW position to transmit. Back to RECEIVE for receiving.

## D. ALIGNMENT AND TROUBLESHOOTING

### GENERAL

The following procedures are given in the order performed during the factory alignment for the transceiver. For home servicing, only partial alignment may be necessary. Read all procedures carefully before commencing either partial or complete alignment. See Figures 4 and 5 for component placement.

#### Equipment Required

1. Calibrated audio frequency signal generator, range 200 to 5000 cps.
2. 500 watt dummy load with output meter
3. Vacuum tube voltmeter
4. Walsco 2543 coil adjustment tool
5. Field strength meter
6. Calibrated RF Signal Generator

#### Pre-Alignment Conditions

1. Neutralizing capacitors C413 set to mid-point and C315 set to approximately 3/4 turn from full compression.
2. Peak 1F transformers for maximum background noise with AF and RF gain full clockwise (either bottom or top core adjustment).
3. Loosely couple field strength meter to C318 (off pin 9 of V4) with alligator clip on ceramic capacitor body.
4. Transmit bias potentiometer full counter-clockwise (maximum bias).

### VFO AMPLIFIER PLATE CIRCUIT ALIGNMENT

With VTVM from pin 1 of V7, Receiver Mixer, to ground, on -15 volt scale, adjust VFO Amplifier Plate coils for peak VTVM heading as follows:

Band	VFO Frequency (kc)	Dial Frequency (kc)	Coil
80	8,975	3,800	L104
40	12,300	7,125	L103
15	16,050	21,225	L102
10	23,325	28,500	L101

### TRANSMITTER MIXER AND DRIVER PLATE CIRCUIT ALIGNMENT

1. Remove screen voltage from V4 and V5 by disconnecting the wire from terminal strip immediately adjacent to V5 base. (A, Fig. 5).

# I MODEL 500 TRANSCEIVER

Operation (Cont.)

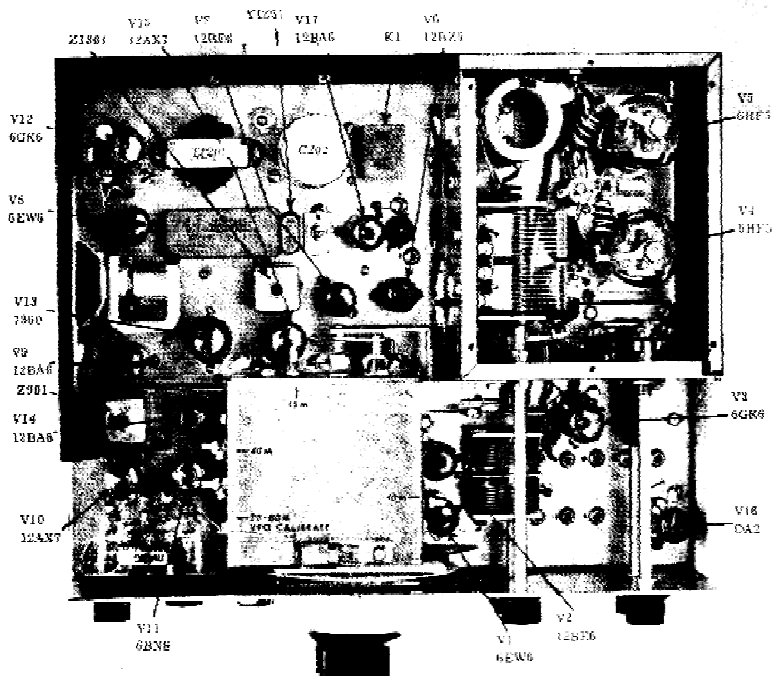
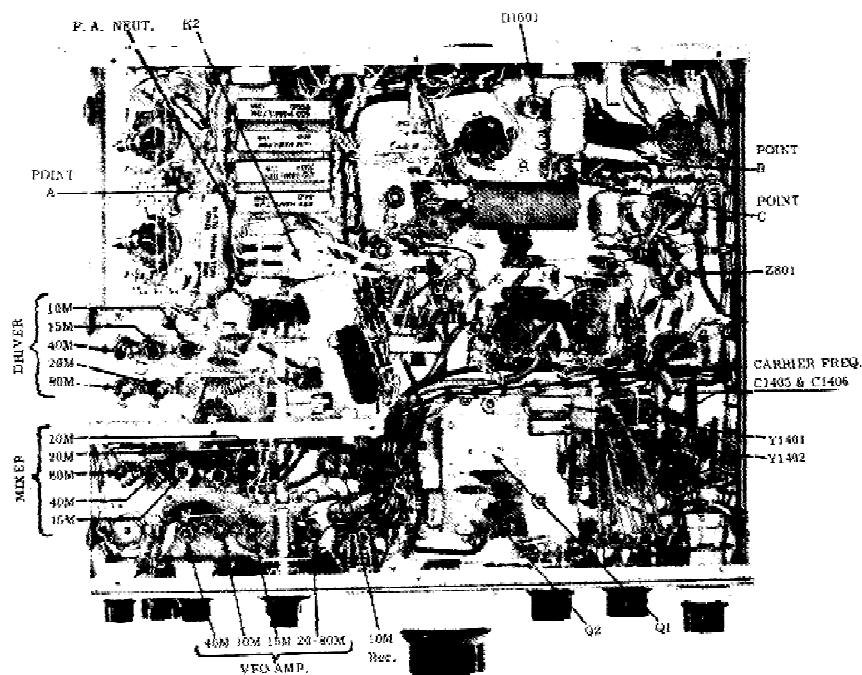


FIGURE 4  
TOP VIEW,  
MODEL 500  
TRANSCEIVER

FIGURE 5  
BOTTOM VIEW,  
MODEL 500  
TRANSCEIVER



2. Connect VTVM across R412, 4.7K resistor between pins 1 and 2 of terminal strip, using 15 volt scale. (Points B and C, Fig. 5).

Procedure:

Adjust band switch and P.A. Grid as shown, and adjust coils for peak VTVM reading as follows, with function switch in Tune position:

Band	P.A. Grid	Dial Freq. (kc)	Adjust
80	12 o'clock	3.800	L205, Z1301 C1402, L1305
40	11 o'clock	7.150	L204, L304
20	11 o'clock	14.150	L203, L303
15	2 o'clock	21.450	L202, L302
10	2:30 o'clock	29.7	L201, L301

\*Note: If VTVM and field strength meter exceed full scale reading, switch to REC. position, actuate push-to-talk circuit, and insert carrier with carrier balance control to keep reading on scale. Field strength meter and VTVM must both peak at same time since it is possible to tune the coils to the VFO frequency on 10 meters. Care must be taken that the coils be tuned properly.

Following the above procedures, replace screen wire to pin 1 of terminal strip adjacent to V5.

#### ALIGNMENT OF 5175 KC FILTER TRAP

With RF and AF gain at midscale, feed 5175 kc signal to antenna connector and adjust L602 until the heterodyne signal is nulled to minimum.

#### ALIGNMENT OF 13 MC FILTER TRAP

Tune VFO to 14,325 kc, insert RF signal to antenna at 13,000 kc and tune RF generator for heterodyne signal in speaker. Adjust L603 and L604 for minimum heterodyne signal.

#### ADJUSTMENT OF CARRIER FREQUENCY

- A. With dummy load and output meter attached, tune transceiver for maximum output.
- B. Null out carrier with PTT pressed and set resting plate current to 50 ma with bias pot.
- C. Connect AF generator to MIC JACK, adjust MIC. GAIN full CCW.

Procedure:

1. With AF generator at 1500 cps, increase MIC. GAIN to produce a 100 ma. reading on the meter.
2. Adjust Z801 for maximum meter reading.
3. Adjust both top and bottom cores of Z1301 for maximum meter reading.
4. Adjust MIC. GAIN for meter reading of 300 ma.
5. Set AF generator to 300 cps. Adjust C1402 for meter reading of 75 ma.

#### P.A. NEUTRALIZATION

With P.A. coarse load in position 1, set freq. to 14.150, PA Plate control at 9 o'clock, insert carrier and peak P.A. Grid control, adjusting Car. Bal. control for 200 MA. Turn PA control slowly through resonance. Cathode current should dip smoothly and rise to 200 MA on the low capacity side of resonance. If, instead, there is a peak above 200 MA either side of the dip, stop rotation of the PA plate control at the peak and adjust C413 to reduce Ip to 200 MA. Repeat above check and readjust as necessary to obtain the desired smooth dip. For 10 meters, use above procedure but adjust #C 315.

#### ADJUSTMENT OF L601

With transceiver tuned to 28.8 mc, and RF and AF gain at maximum, adjust L601 for maximum background noise.

#### S-METER ADJUSTMENT

With antenna disconnected and with RF Gain fully clockwise, set R605, located on rear panel, for zero meter reading. Make sure no local signals are being received. It will be noticed that a slight change in S-meter zero setting will change when switching from band to band. Also, when using the ANL circuit there will be a decrease in the S-Meter reading. This is normal.

#### VFO ALIGNMENT

A locking type trimmer condenser is provided for each VFO range, of which there are four. The same range is used on both 20 and 80 meters. Dial tracking has been factory set by pruning the coil, and will not ordinarily require further adjustment.

When dial calibration changes beyond the adjusting range of the front panel dial set control, calibration may be restored by carefully adjusting the trimmer for that range. It may be necessary to loosen the locknut. Be sure to tighten it again.

The following chart lists the actual oscillating frequency of the VFO at band edges:

Dial Frequency	Oscillator Frequency
3500 (13.85)	8673 KC
3800 (14.15)	8973
4000 (14.35)	9173
7000	12,173
7200	12,373
7300	12,473
21,000	15,827
21,250	16,077
21,450	16,277
28,000	22,827
28,500	23,327
29,000	23,827
29,700	24,527

TROUBLESHOOTING GUIDE	
DEFECT	POSSIBLE CAUSE
PA IDLING CURRENT UNSTABLE	1. Defective 6HF6 — See NOTE. 2. Defective Bias Potentiometer 3. Defective Bias Supply
INABILITY TO LOAD TO 400-500 MA. (SEE PAGE 10)	1. PA Grid Improperly Tuned 2. Bandswitch Improperly Set 3. Antenna Not Resonant at Frequency 4. Defective Transmission Line 5. Defective Mobile Antenna Coil 6. V2, V3, V4, V5 Defective 7. R407 or R408 Defective
INSUFFICIENT CARRIER SUPPRESSION	1. Carrier Balance Control Improperly Adjusted 2. Defective 5950 Balanced Modulator 3. Carrier Oscillator Frequency Incorrect
INSUFFICIENT SIDEBAND SUPPRESSION	1. Excessive M.C. Gain 2. Incorrect PA Load Adjustment 3. Carrier Oscillator Frequency Incorrect
MICROPHONICS IN RECEIVER	1. Z501 Improperly Tuned 2. V14, V10, V8, V7, or V6 Defective
LOW RECEIVER SENSITIVITY	1. PA Grid, Plate, or Load Improperly Set 2. Bandswitch Improperly Set 3. K2 Back Contacts Defective 4. V6, V7, V8, V9, V10, V11, V12 Defective

NOTE: It is recommended that the final tubes be replaced with General Electric tubes for neutralization purposes.

TUBE TYPE		VOLTAGE CHART											
		PIN NO.											
		1	2	3	4	5	6	7	8	9	10	11	12
V1 6EW6	R	0	-75	8.3	0	130	130	0					
	T	0	-75	8.3	0	135	135	0					
V2 12BH6	R	-90	0	12.6	0	245	340	360					
	T	-2.0	0	12.6	0	24.5	100	0					
V3 6GX6	R	0	-30	0	6.3	0	0	255	265	0			
	T	0	-6.0	0	6.3	0	0	250	260	0			
V4 9HF6	R	6.3	—	—	—	-75	0	—	0	-75	0	—	0
	T	6.3	—	—	-12	-75	215	—	—	-215	-75	12	—
V5 6HF5	R	12.6	—	—	—	0	-75	0	—	0	-75	0	—
	T	12.6	—	—	-12	-75	215	—	—	-215	-75	12	—
V6 12B26	R	-06	8	0	6.3	230	140	0					
	T	-1	0	0	6.3	230	20	0					
V7 12B28	R	-30	0	12.6	0	235	75	—					
	T	-3.0	0	12.6	0	210	20	3					
V8 6EW6	R	-1	-5	12.6	6.3	230	130	0					
	T	-1	-0.5	12.6	6.3	205	130	0					
V9 12BA6	R	-06	0	12.6	0	230	105	8.2					
	T	-6	0	12.6	0	220	75	0					
V10 12AX7	R	110	-95	1	12.6	0	180	0	1.5	6.3			
	T	40	-25	-5	12.6	0	155	100	5	6.3			
V11 6BN8	R	0.2	2.6	0.2	12.6	6.3	0	200	4	70			
	T	0	2.3	0	12.6	6.3	3	160	2	40			
V12 6GK5	R	0	6	—	12.6	6.3	—	240	230	0			
	T	0	-21	—	12.6	6.3	—	260	250	0			
V13 7360	R	0	0	90	6.3	0	225	225	25	23			
	T	0	70	0	6.3	0	135	135	23	23			
V14 12BA6	R	-7	0	0	12.6	60	100	1					
	T	-7	0	0	12.6	60	100	1					
V15 12AX7	R	95	0	0.5	—	—	—	70	0.2	0			
	T	60	0	0.3	6.3	-6.3	90	0.2	0	0			
V16 0A2	R	150	—	—	—	—	—	160	0				
	T	150	—	—	—	—	—	160	0				
V17 12BA6	R	0	0	12.6	0	240	240	70					
	T	0	0	12.6	0	240	240	70					

All Voltage Measurements Made With Simpson 260, 20 K Ohm per Volt, Or Equivalent

# E. PARTS LIST

## CAPACITORS

C101 .002, 20% 1 KV Disc  
 C102 .01, +80-20%, 500V Disc  
 C103 .002, 20% 1 KV Disc  
 C104 .01, +80-20%, 500V Disc  
 C105 .01, +80-20%, 500V Disc  
 C106 50, 5% 500V Mica  
 C201 .01, +80-20%, 500V Disc  
 C202 40-30-10-20 Mfd.  
 450-450-450-25 WV Elect.  
 C203 .002, 20% 1 KV Disc  
 C204 .001, 5% 500V Mica  
 C205 82, 5% 500V Mica  
 C206 24, 5% 500V Mica  
 C207 120, 5% 500V Mica  
 C208 27, 5% 500V Mica  
 C209 20, 5% 500V Mica  
 C210 27, 5% 500V Mica  
 C211 50, 5% 500V Mica  
 C212 1.5-20pf Mica Trimmer  
 C213 3.3 pf 10% 500V Ceramic  
 C214 .1 mf 10% 200V Mylar  
 C215 100, 5% 500V Mica  
 C216 .5, 10% NPO Disc  
 C217 10, 5% 500V Mica  
 C218 100, 5% 500V Mica  
 C219 1.5-20 pf Mica Trimmer  
 C301 .01 +80-20%, 500V Disc  
 C302 .01 +80-20%, 500V Disc  
 C303 100 5% 500V Mica  
 C305 100, 5% 500V Mica  
 C306 27, 5% 500V Mica  
 C307 50, 5% 500V Mica  
 C308 .002, 20% 1 KV Disc  
 C309 510, 5% 500V Mica  
 C310 540, 5% 500V Mica  
 C311 100, 5% 500V Mica  
 C312 680, 5% 500V Mica  
 C313 50, 5% 500V Mica  
 C314 91, 5% 500V Mica  
 C315 1.5-20 pf Mica Trimmer  
 C316 15, 20% 3KV Disc  
 C317 330, 5% 500V Mica  
 C318 .002, 20% 1 KV Disc  
 C319 68pf DM15  
 C320 1.5-20pf Mica Trimmer  
 C401 .002, 20% 1 KV Disc  
 C402 .002, 20% 1 KV Disc  
 C403 .01, +80-20%, 500V Disc  
 C404 .01, +80-20%, 500V Disc  
 C405 .01, +80-20%, 500V Disc  
 C406 10 mf 150 WV Electrolytic  
 C407 .01, +80-20%, 500V Disc  
 C408 .01, +80-20%, 500V Disc  
 C409 .01, +80-20%, 500V Disc  
 C410 .002, 1 KV Disc  
 C411 .002, 20% 2 KV Disc  
 C412 .002, 20% 2 KV Disc  
 C413 20 pf P.A. Neut. Trimmer  
 C414 15, 20% 3 KV Disc  
 C415 270, 5% 2500V Mica

C416 270, 5% 2500V Mica  
 C417 320 pf P.A. Tune  
 C418 50 10% 6 KV Disc  
 C419 100 10% 6 KV Disc  
 C420 410 pf P.A. Fine Load  
 C421 Two 150 = 5% 1000 WV Mica  
 C422 Two 150 = 5% 1000 WV Mica  
 C423 330, 10% 500V Mica  
 C424 330, 10% 500V Mica  
 C425 330, 10% 500V Mica  
 C426 330, 10% 500V Mica  
 C427 330, 10% 500V Mica  
 C428 330, 10% 500V Mica  
 C429 330, 10% 500V Mica  
 C430 .002, 20% 1 KV Disc  
 C431 .47 10% 100V Mylar  
 C432 .01 Disc SM  
 C601 .01 +80-20%, 500V Disc  
 C602 .01 +80-20%, 500V Disc  
 C603 .01 +80-20%, 500V Disc  
 C604 10, 5% 500V Mica  
 C605 5, 10% NPO  
 C606 100, 5% Mica  
 C607 1.5-20 pf Mica Trimmer  
 C701 30, 10% 1 KV Disc  
 C702 30, 10% 1 KV Disc  
 C703 .01 +80-20% 500V Disc  
 C704 220 5% 500V Mica  
 C705 430 5% 500V Mica  
 C709 .01, +80-20% 500V Disc  
 C802 .01, +80-20% 500V Disc  
 C803 .01, +80-20% 500V Disc  
 C804 10, 10% 1 KV Disc  
 C805 .01, +80-20% 500V Disc  
 C806 50 5% N750 Disc  
 C807 50 5% N750 Disc  
 C901 .01 +80-20% 500V Disc  
 C902 .01 +80-20% 500V Disc  
 C903 .01 +80-20% 500V Disc  
 C904 .01 +80-20% 500V Disc  
 C1001 100, 5% 500V Mica  
 C1002 220, 20% 1 KV Disc  
 C1003 .002, 20% 1 KV Disc  
 C1004 .001, 20% 1 KV Disc  
 C1005 .001, 20% 1 KV Disc  
 C1006 .22 mf 200VDC  
 C1007 500, 20% 1 KV Disc  
 C1008 .002 20% 1 KV Disc  
 C1009 .002 20% 1 KV Disc  
 C1101 .001 20% 1 KV Disc  
 C1102 .01 +80-20% 500V Disc  
 C1104 .001 500V Disc  
 C1201 .01 +80-20% 500V Disc  
 C1202 220 20% 1 KV Disc  
 C1203 .0047 10% 1000V Mylar  
 C1204 .1 200VDC  
 C1301 .01 +80-20% 500V Disc  
 C1302 220 20% 1 KV Disc  
 C1303 .01 +80-20% 500V Disc  
 C1304 .002 20% 1 KV Disc  
 C1305 .01 +80-20% 500V Disc

C1306 .01 +80-20% 500V Disc  
 C1401 27, 5% 500V Mica  
 C1402 50, 5% 500V Mica  
 C1403 50, 5% 500V Mica  
 C1404 15, 5% 500V Mica  
 C1405 6-30 pf Ceramic Trimmer  
 C1406 6-30 pf Ceramic Trimmer  
 C1407 .01 +80-20% 500V Disc  
 C1501 .01, 400V Mylar  
 C1502 .01 +80-20% 500V Disc  
 C1503 .01 +80-20% 500V Disc  
 C1504 100, 20% 1 KV Disc  
 C1505 .01 +80-20% 500V Disc  
 C1506 20 mf 25VDC  
 C1507 .47 200V  
 C1601 80 mf, 150 WV Electrolytic  
 C1602 .01 +80-20% 500V Disc  
 C1603 1 MF 50V  
 C1604 2 MF 50V  
 C1701 270, 2% SM  
 C1702 470, 5% SM  
 C1703 430, 2% SM  
 C1704 27, 5% SM  
 C1705 430, 2% SM  
 C1706 5-12 Main Tuning  
 C1707 .01, 500V Disc  
 C1708 5 pf Trimmer  
 C1709 22, N 220 Disc  
 C1710 10, N 150 Disc  
 C1711 5, N 150 Disc  
 C1712 5 pf Trimmer  
 C1713 10, N 150 Disc  
 C1714 25 NPO Disc  
 C1715 5 pf Trimmer  
 C1716 .01 500V Disc  
 C1717 2 pf Dial Set  
 C1718 22, N 75 Disc  
 C1719 39 NPO Disc  
 C1720 5 pf Trimmer  
 C1721 5, N 75 Disc  
 C1751 5-80 Ceramic Trimmer  
 C1752 150, 5% 500V Mica  
 C1A-B 85 pf per section

## RESISTORS

R101 82 ohms  
 R102 56 ohms  
 R103 47K - 1 watt  
 R104 12K - 2 watt  
 R201 27K  
 R202 18K - 2 watt  
 R203 4.7K - 1 watt  
 R204 6.8K  
 R205 6.8K  
 R206 4.7K  
 R207 27K  
 R301 100K  
 R302 270K  
 R303 100 ohms  
 R304 10K

R305 8.2K  
 R306 10K  
 R401 2.2 Meg  
 R402 1K  
 R403 100 ohms  
 R404 100 ohms  
 R405 Selected  
 R406 470 - 5% - 1/2 W  
 R407 10K - 10 watt  
 R408 1 ohm - 5% - 1 watt  
 R409 1 ohm - 5% - 1 watt  
 R410 10K Bias Pot.  
 R411 10K - 1 watt  
 R412 4.7K  
 R413 4.7K  
 R601 1K  
 R602 56 ohms  
 R603 47K - 1 watt  
 R604 1K  
 R701 27K  
 R702 22K - 1 watt  
 R801 470 ohms  
 R802 56 ohms  
 R803 47K  
 R804 1K  
 R902 100 ohms  
 R903 27K - 1 watt  
 R904 47 ohms  
 R905 100K  
 R906 100K - 5%  
 R907 1K SMeter Zero  
 R1001 10K  
 R1002 1K  
 R1003 100K  
 R1004 270K  
 R1005 2.2 Meg.  
 R1006 2.2 Meg.  
 R1007 270K  
 R1008 1 Meg.  
 R1009 2.7K  
 R1010 100K  
 R1101 470K  
 R1102 1K  
 R1103 47K  
 R1104 2.7K  
 R1105 270K  
 R1106 120, 1 watt  
 R1201 1 Meg. A.F. Gain Pot.  
 R1202 470K  
 R1203 270K  
 R1204 470K  
 R1301 47K  
 R1302 47K  
 R1303 150K  
 R1304 4.7K  
 R1305 5K CAR. BAL. Pot.  
 R1306 47K  
 R1307 47K  
 R1308 100K  
 R1309 100K  
 R1310 27K  
 R1401 1 Meg.  
 R1402 270 ohm  
 R1403 27K

R1404 27K  
 R1405 1K  
 R1501 150K  
 R1502 1K  
 R1503 1 Meg. MIC GAIN Pot.  
 R1504 270K  
 R1505 2.2 Meg.  
 R1506 47K  
 R1601 800 - 10 watt  
 R1602 900 - 10 watt  
 R1603 27K  
 R1604 27K  
 R1605 750 - 10 watt  
 R1606 6K - 10 watt  
 R1608 100K  
 R1609 470K  
 R1610 270K  
 R1611 270K  
 R1612 10K RF Gain Pot.  
 R1613 270K  
 R1614 27K  
 R1701 1K, 5%  
 R1702 1K, 5%  
 R1703 1K, 5%  
 R1704 470 ohms 5%  
 R1705 2.7K  
 R1706 2.7K  
 R1707 470 ohms 5%  
 R1751 1 Meg.  
 R1752 27K  
 R1753 100K

#### COILS

L101 23 mc - 2 uh  
 L102 16 mc - 4 uh  
 L103 12 mc - 7 uh  
 L104 9 mc - 4 uh  
 L201 28 mc - 2 uh  
 L202 21 mc - 2 uh  
 L203 14 mc - 3.2 uh  
 L204 7 mc - 3.6 uh  
 L205 4 mc - 11 uh  
 L206 RFC - 200 uh  
 L207 5173 kc Trap - Toroid  
 L301 28 mc - 2 uh  
 L302 21 mc - 2 uh  
 L303 14 mc - 3.2 uh  
 L304 7 mc - 3.6 uh  
 L305 4 mc - 11 uh  
 L306 RFC - 200 uh  
 L401 14 mc - 0.8 uh  
 L402 4 mc - 6 uh  
 L403 RFC - 38 uh  
 L404 RFC - 200 uh  
 L405 RFC - 55 uh  
 L406 RFC - 17 uh  
 L601 28 mc - 1.2 uh  
 L602 5175 kc - 90 uh  
 L603 13 mc. 30 uh  
 L604 13 mc. 1.5 uh  
 L605 18 mc. 17 uh  
 L701 RFC - 200 uh  
 L1001 RFC - 200 uh  
 L1701 80/20M - VFO Coil

L1702 40M - VFO Coil  
 L1703 15M - VFO Coil  
 L1704 10M - VFO Coil  
 L1705 RFC - 200 uh  
 L1706 RFC - 200 uh

#### TRANSFORMERS

Z301 Parasitic Suppressor  
 Z401 Parasitic Suppressor  
 Z501 Parasitic Suppressor  
 Z801 5175 kc I.F. Trans.  
 Z901 5175 kc I.F. Trans.  
 Z1301 5175 kc BAL. MOD. Trans.  
 T1201 A.F. Output Trans.

#### SWITCHES

S-1 Power On-Off  
 (part of AF & RF Gain Control)  
 S-2 CAL-REC-TUNE-CW  
 S-3 VOX PTT  
 S-4 A-B-C-D-E-F Bandswitch  
 S-5 PA Coarse Load  
 S-6 Sideband Selector  
 S-7 ANL ON-OFF Switch  
 S-46 OSC. Selector

#### DIODES

D401 TS-2 ALC Diode  
 D402 TS-2ALC Diode  
 D1001 1N 34A ANL  
 D1002 1N 34A ANL  
 D1601 1N 2974A Zener  
 D1602 TS-2 Relay Quieting  
 D1603 TS-2 AGC Charging bypass

#### RELAYS

K1 4PDT Relay, 12 VDC Coil  
 K2 PDT Relay, 12 VDC Coil

#### CRYSTALS

Y1401 5176.8 KC Car. Osc.  
 Y1402 5172.8 KC Car. Osc.  
 Y1751 100 KC Crystals Calibrator

#### TUBES

V1 6EW6 VFO Amplifier  
 V2 12BE6 Trans. Mixer  
 V3 6GK6 P.A. Drive  
 V4 6HF5 Power Amplifier  
 V5 6HF5 Power Amplifier  
 V6 12BZ6 Rec. R.F. Amp.  
 V7 12BE6 Rec. Mixer  
 V8 6EW6 1st I.F. Amp.  
 V9 12BA6 2nd I.F. Amp.  
 V10 12AX7 Prod. Det./Rec. A. F.  
 V11 6BN8 AGC Amp./Rect.  
 V12 6GK6 A.F. Output Amp.  
 V13 7360 Bal. Mod.  
 V14 12BA6 Carrier Oscillator  
 V15 12AX7 Mic. Amplifier  
 V16 OA12 Voltage Regulator  
 V17 12VA6 100 kc Oscillator

#### TRANSISTORS

Q1 2N706  
 Q2 2N706



